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Analysis and Design of Low Pass Filter by Using DGS for WLAN Application

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ABSTRACT: The performance of Butterworth low pass filter (LPF) with defected ground structure (DGS) is studied and simulated. Calculation and comparison of the response of low pass filter (LPF) with and without using defected ground structure (DGS) was done. The proposed DGS cell provides low-pass characteristic whose cutoff frequency could be changed by tuning its dimensions. Results are simulated using computer simulation technology software (CST). In this paper, 5th order stepped impedance low pass microstrip line filter have been designed at 5.6 GHz frequency and implemented on FR4 substrate of relative permittivity is 4.3. The undesired sidebands and fluctuations of response are reduced by using defected ground structure (DGS).

KEYWORDS: Low Pass Filter (LPF), Defected Ground Structure (DGS), Microstrip Stepped Impedance Filter, Butterworth Low Pass Filter.

I. INTRODUCTION

Defected ground structure for microstrip line was most common topic for research at recent year. They are giving a lot of different structure for implementing DGS [1]. Microwave filter designs have been at the forefront of research in both industry as well as in academia due to increasing specification levels and demand for advanced communication systems.

In addition to PBG (photonic bandgap) and EBG (electromagnetic bandgap) structure, DGS was created by etching different shapes in ground plane. Which increase the inductance and capacitance values of microstrip line, so undesired output response fluctuations will be eliminated and the output is sharp stopband in case of LPF [2] and increased bandwidth in case of BSF [3]. DGS has property of neglecting electromagnetic wave in certain frequency and direction, and most important function of these structures is the filtering of frequency bands, and harmonics of the filter in microwave circuit.

The realizable filters that are in common use are Butterworth filter, Chebyshev filter and Bessel filter. Utilization of Butterworth low pass filter exist contradiction problems between test precision, stability and response time. Low order Butterworth low pass filter has rapid response, small overshoot but bad in test precision, while high order Butterworth low pass filter, is good in test precision but has slow response with large overshoot and poor stability [4-5]. The type of construction of this filter is a reflective filter which is consists of capacitive and inductive elements producing ideally the zero reflection loss in the pass band region and very high attenuation in the stop band region [6]. The practical filters have small non-zero attenuation in the pass band, a small signal output in the attenuation or stop band due to the presence of resistive losses in reactive elements of propagating medium [7].

II. FILTER DESIGN PROCEDURE

In stepped impedance LPF and L-C ladder type LPF with open circuited stub having frequencies of infinite attenuation at f = infinite. To find sharper cutoff for a given number of reactive elements it is desirable to use filter structure giving infinite attenuation at finite frequencies.

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Figure 1 shows a general structure of the stepped-impedance low pass microstrip filters which use a cascade structure of alternating high and low impedance transmission lines. These are much shorter than the associated guided-wavelength, so as to act as semi lumped elements [8]. The high-impedance lines act as series inductors and the low-impedance lines act as shunt capacitors. Therefore, this filter structure is directly realize the L-C ladder type of lowpass filters of figure 2.

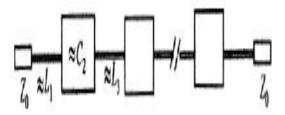


Figure 1 General structure of the stepped-impedance low pass microstrip filter.

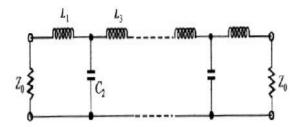


Figure 2 L-C ladder type of low pass filter.

LPF was design at the cut off frequency of f_c=5.6GHz and formula which is used for the design of LPF is

Synthesis of W/h

$$\frac{W}{h} = \frac{8 e^A}{e^{2A} - 2}$$

With

$$A = \frac{Z_c}{60} \left[\frac{\varepsilon_r + 1}{2} \right]^{0.5} + \frac{\varepsilon_r + 1}{\varepsilon_r + 1} \left[0.23 + \frac{0.11}{\varepsilon_r} \right]$$

Where $Zc=Zo=50\Omega$ and ε_r (dielectric constent) = 4.4, W= width, h= height of dielectric which is taken as 1.6mm.

Effective dielectric constant of dielectric material given by equations

For $W/h \le 1$

$$\varepsilon_{re} = \frac{\varepsilon_r + 1}{2} + \frac{\varepsilon_r - 1}{2} \left(1 + 12 \frac{h}{w}\right)^{-0.5}$$



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For W/h>1

$$\varepsilon_{re} = \frac{\varepsilon_r + 1}{2} + \frac{\varepsilon_r - 1}{2} \left[\left(1 + 12 \frac{h}{w} \right)^{-0.5} + 0.04 \left(1 - \frac{w}{h} \right)^2 \right]$$

Whereas guided wavelength is given by equation given below

$$\lambda_{\rm g} = \frac{300}{f({\rm GHz})\sqrt{\epsilon_{\rm re}}}$$

 ε_{re} = Effective dielectric constant

Values of inductor and capacitor are given by

$$C_{i} = \frac{1}{2\pi Z_{0} f_{c}} g'_{i} L_{i} = \frac{1}{2\pi f_{c}} Z_{0} g_{i}$$
For $i = 1, 2, 3$

Calculation of length of inductor and capacitor is done using formula

$$l_{Li} = \frac{\lambda_{\text{gL}}(f_{\text{c}})}{2\pi} \sin^{-1}\left(2\pi f_{\text{c}} \frac{L_{\text{i}}}{Z_{\text{oc}}}\right)$$
$$l_{\text{Ci}} = \frac{\lambda_{\text{gc}}(f_{\text{c}})}{2\pi} \sin^{-1}(2\pi Z_{\text{oc}}C_{\text{i}})$$

III. IMPLEMENTATION OF $5^{\text{\tiny TH}}$ ORDER STEPPED IMPEDANCE BUTTERWORTH FUNCTION LOW PASS FILTER

The design of low pass filters involves two main steps. The first one is to select an appropriate low pass prototype. The choice of the type of response, including Pass band ripple and the number of reactive elements will depend on the required specifications.

Element values considered stepped impedance Butterworth function low pass prototype filter source resistance $g_0=g_6=1.0$, cutoff frequency $\Omega_c=1 (rad/s)$, pass band ripple $L_{Ar}=3.01dB$, $g_1=0.618$, $g_2=1.618$, $g_3=2$, $g_4=1.618$, $g_5=0.618$. Where g_i represent inductance of the inductor (for i=1,3,5) and g_i represent capacitance of the capacitor (for i=2,4).

After that transformed to the L-C elements for the desired cutoff frequency and the required source impedance, which is normally 50 ohms for microstrip filters. The next main step in the design of microstrip low pass filters [1] is to find an appropriate micro strip realization that approximates the lumped element filter [9-10].

The dimensions of the length and width of inductor (in mm) and capacitor (in mm) with their respective inductor and capacitor values are shown in the table 1.

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Table1. Microstrip design parameters for 5th order stepped impedance Butterworth low pass filter.

S. No.	Values of impedances (ohm)	Values of inductor(nH) and capacitor(pF)	Length and width of inductor and capacitor (mm)
1	Z _{0L} =93, Z _{0c} =24	L ₁ =3.94, C ₂ =0.703, L ₃ =6.51, C ₄ =0.703, L ₅ =3.94	$l_1 = 10.79, l_2 = 11.96, l_3 = 23.4, l_4 = 11.96, $ $l_5 = 10.79, w_L = 0.652, w_C = 8.82$
2	Z ₀ =50		l ₀ =5, w ₀ =3.058

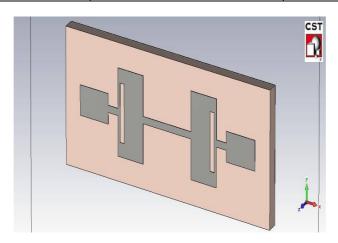


Figure 3 (a) Front view of proposed designed microstrip stepped impedance butterworth function LPF.

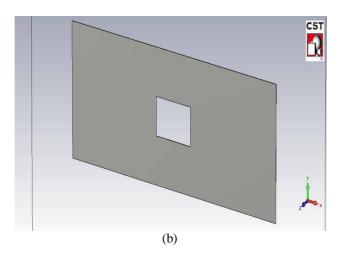


Figure 3 (b) Back view of ground structure of the designed microstrip stepped impedance function LPF.

The proposed design of the 5th order stepped impedance Low Pass Filter is shown in the figure 3. FR4 lossy material with dielectric constant of 4.4, substrate height of 1.6mm and loss tangent 0.02 was used in the designing of low pass filter a rectangular shaped slot of dimension 6mm X 4mm is introduced in the centre of the ground plane which shows the ground as defected.

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One rectangular shaped slot of equal size is introduced in the both capacitor part ($C_2 \& C_4$) of the proposed design of microstrip Butterworth low pass filter. Dimension of rectangular shaped slots is 0.7mm X 6mm and it is placed form the 0.4mm far away from the inductor part ($L_1 \& L_5$) of the design.

IV. SIMULATION RESULTS

The simulated results of the stepped impedance Butterworth Low Pass Filter for WLAN application is shown in the figure 4. The graph obtains after the simulation in CST Software [11].

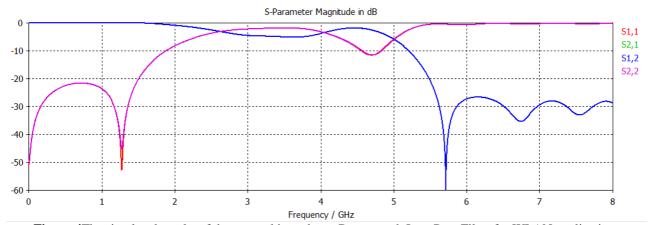


Figure 4The simulated results of the stepped impedance Butterworth Low Pass Filter for WLAN application

From the response shown in figure 4, it is clear that the cut-off frequency is found to be 5GHz for stepped-impedance low pass filter. Hence stepped impedance low pass filter is capable of passing the frequency less than 5GHz & reject the frequency after 5GHz.



Figure 5 Amplitude versus frequency response of the stepped impedance Butterworth Low Pass Filter for WLAN application.

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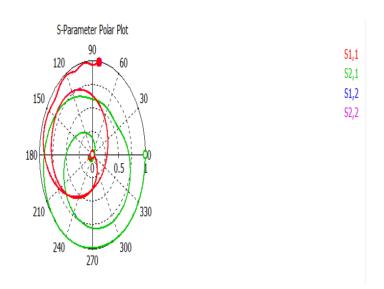


Figure6The angular plot of the result of the stepped impedance Butterworth Low Pass Filter for WLAN application

Table 2. Comparison of Butterworth filter parameters

Parameters	Navita Singh and Dr. Avinash Kumar 2011	Dhirendra kumar and Ashok De 2011	This work
Cutoff frequency	1.5 GHz	1.7 GHz	5 GHz
Insertion loss	-33	-22.9	-60
Directivity	-	-	4.874
Gain	3 dB	-	3.526 dB
Efficiency	-	-	11.55 dB

V. CONCLUSION

The proposed design was implemented and analyzed at the centre frequency 5GHz. A sharp rate of cutoff with reduce label of sideband fluctuation of the response achieved by introducing the slots in the ground plane structure which is behave as defected and two equal slots of rectangular shaped in the structure of low pass filter I have made slots in the ground structure produces so that the low power will consumption takes places and use the dielectric FR4 0.038mm and make the circuit as an ideal and passes the most of the signal at the desires frequency and the graphical structure is maximally flat and the try to make the minimum insertion loss -60dB which is very less loss and the efficiency is 11.55dB and the gain is 3.526dB.

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