

(An ISO 3297: 2007 Certified Organization)

Vol. 4, Issue 7, July 2015

Design of BIST Architecture of 8×8 SRAM Testing using Transient Current Method

Rohit Jain¹, Rajesh Khatri²

PG Student [Microelectronics & VLSI Design], Dept. of E&I Engineering, SGSITS Indore, MP, India¹

Assistant Professor, Dept. of E&I Engineering, SGSITS Indore, MP, India²

ABSTRACT: BIST Architecture is proposed in this paper for testing of 8×8 SRAM based on transient current method. Basically in this method comparison is made between the faulty SRAM & non faulty SRAM by conversing transient current into voltage. The fault is detected by mismatched waveform of transient voltage. This method is faster than the March Algorithm method. In transient current method only 2 write operations are required instead of 4 as in March Algorithm method. Transient current method is low power technology. All the simulations i.e. write operation, transient current and transient voltage is carried out on 6T SRAM cell. Simulations are carried out on 8×8 SRAM circuit & are designed on Cadence UMC 180nm process technology file with supply voltage 2.5V. Cadence specter simulator is used to simulate proposed BIST.

KEYWORDS: SRAM, March Algorithm, IDDT current, Vddt Sensor, Decoder, BIST.

I.INTRODUCTION

Memories are important part of System on chip (SOCs). The forecast for 2013 from International Technology Roadmap for semiconductors (ITRS) [1][2] states that the 90% of the area of SOCs will be made up of memories, mostly static Random Access Memory. To improve the system performance large array SRAMs is very helpful. In SRAM testing various fault models such as transition fault, Open defect fault, Bridging fault and Stuck at fault needed to be used. To detect these fault March Algorithm [3][4] is widely used. But it is time consuming process. March Algorithm requires 4 write operations, but IDDT method takes 2 write operations. So that IDDT current method is faster than March Algorithm. In Transient current method there are many parameters of waveform that may be change. In transient current testing, there are usually measured parameters (Fig 1), which include current waveform width, the peak value of waveform. Testing using quiescent current (IDDQ) is also used [5]. Some defects in SRAM may not be detected by using IDDQ. But all open defect fault is detected in SRAM cell by monitoring a transient current pulse during a write or read operation.

In this paper transient current method is used in order to experimentally analyze electrically defects in SRAMs. All possible defects in the memory cell array will be given, together with a systematic way of reducing the number of simulated defects. SPICE simulation will be performed, and the behaviours will be transformed into functional fault models.

This paper is organised as follows: section II established designing of SRAM cell, section III describe, how to transient current flow in SRAM cell and types of faults, section BIST architecture of SRAM is designed, section V show the faults in SRAM based on simulation result, section VI ends with conclusion.

II.BACKGROUND

A. SRAM Cell:

A low power SRAM cell may be designed simply by cross coupled CMOS inverter instead of resistive load NMOS inverters. In this case the standby power consumption of the memory cell will be relatively low due to small leakage current. The SRAM cell design consisted of sizing the transistor for read and write stability. The layout was performed to create as compact a cell as possible. Accordingly, a 6 transistor (6T) cell is shown in figure 1. The SRAM design consisted of sizing the transistors and determining the read and write stability. The layout was performed to create as compact a cell possible [6].

For write stability voltage across v_2 of M_4 is less than threshold voltage (v_{th}) of M_4 . Bit Line (BL) is pulled high to write a '1' into the 6T cell. The exact size of M_4 and M_6 can be determined from the pull up ratio (PR).



(An ISO 3297: 2007 Certified Organization)

Vol. 4, Issue 7, July 2015





Where,

$$PR = \frac{W6/L6}{W4/L4}$$

If L₄=L6=180nm

Then,

PR has to be at least less than 1.8 to ensure the write stability. A PR value of 1 is chosen for design of 6T cell. Then

 $PR = \frac{W6}{W4}$

$$W_4 = W_6 = 240 nm$$

For read stability the voltage v_1 across M_1 is less than the threshold voltage (=0.4v). It can be chosen by the size of M_1 and M_3 . The exact size of $M_1 \& M_3$ is determined by Cell Ratio (CR).

Where,

$$CR = \frac{W1/L1}{W3/L3}$$

If $L_1 = L_5 = 180$ nm

Then,

i.e.

$$CR = \frac{W1}{W3}$$

CR has to be greater than 1.2 to ensure the read stability. A CR value of 1.2 is chosen for design of 6T SRAM cell.

$$W1 = 360$$
nm, $W5 = 240$ nm.



(An ISO 3297: 2007 Certified Organization)

Vol. 4, Issue 7, July 2015

B. Transient current:

Basically transient current flow in SRAM, when direct current flow between V_{dd} and ground in combination with the charging and discharging of node capacitance.

Now consider the write "0" operation, assuming that logic "1" is stored in the SRAM cell initially. Figure shows the voltage levels in the CMOS SRAM cell at the beginning of the data-write operation. The transistors M_1 and M_6 are turned off, while the transistors M_2 and M_5 operate in the linear mode. The current through M_5 and M_2 is transient current. These current directly flow from V_{dd} to ground. This is shown in figure 2.



Figure 2: Transient current in SRAM cell [7]

C. Fault introduced:

The Fault occurs in SRAM cell due to logical or electrical design errors and faults in fabrication due to stressful operating conditions, namely reliability problem. Faults in fabrication are fault that were not intended. The faults in fabrication are occurring due to mistake in fabrication process or designing problem. The testing mechanism is based on comparison between logical behaviour of good SRAM cell and faulty SRAM cell. The fault in SRAM cell occurs due to open and bridging fault shown in figure 3.



Figure 3: Open and bridging fault in SRAM cell [8]

Due to different values of resistances with different nodes, open and bridging faults are present in SRAM cell. Table I shows open and bridging faults by each resistor.



(An ISO 3297: 2007 Certified Organization)

Vol. 4, Issue 7, July 2015

Resistance	Resistance value	Name of fault	Fault model
R ₁	1MΩ	Open defect 1	TF
R_2	1MΩ	Open defect 2	DRF
R ₃	1MΩ	Open defect 3	S-a-1
R_4	1MΩ	Open defect 4	DRF
R ₅	1MΩ	Open defect 5	SOF
R ₆	1MΩ	Open defect 6	DRF
R ₇	10Ω	Bridging defect 1	SAF
R ₈	10Ω	Bridging defect 2	S-a-1
R ₉	10Ω	Bridging defect 3	S-a-0
R ₁₀	10Ω	Bridging defect 4	CF
R ₁₁	10Ω	Bridging defect 5	CF

Table I Resistance introduced to fault model

SAF: Stuck at fault occure in SRAM cell, when it has always zero or has always one. Stuck at faults are represents by S-a-0 and S-a-1.

TF: Transition fault in SRAM, when it makes a transition 0-1 or 1-0 but fails to other direction.

SOF: Stuck open fault occure in SRAM cell when it has the unexpected result, i.e. niether write nor read operation.

CF: Coupling fault in SRAM cell is considered, when write operation to the cell affects the contents of nieghbouring cell vice versa.

DRF: Data retaction fault is occure when it loses its content after certain period.

IV. BIST ARCHITECTURE

A. Vddt sensor:

IDDT current [9][10] is very fast action. It is extremely difficult to sense and process it. Mostly in low power technologies, processing the dynamic supply current is almost insoluble. Thus by transforming the current to voltage, and handling the resulted voltage waveform is a possible solution.

The Vddt sensor has a PMOS transistor, which have grounded gate, which acts like a low value resistance. 1pf value of capacitor is used; the Vddt sensor is shown in figure 4. M_2 and M_5 act as two emitter follower, which have same DC condition.

Then current through M_2 and M_5 are also identical. Here capacitor C is used to stretch the voltage at V (SENSE), This is appear in wave form with small spikes (μv). This voltage is sent to the differential pair amplifier. The output of differential amplifier is applied to operational amplifier. The operational amplifier transform the μv to millivolt or volt.



(An ISO 3297: 2007 Certified Organization)

Vol. 4, Issue 7, July 2015



Figure 4: Vddt sensor

In Vddt sensor all transistors should be in saturation, sizing of transistor shown in given table III.

Transistor	Width of transistor	Length of
	(W in µm)	transistor (L in µm
)
M_1	10	180
M_2	8	180
M ₃	6	180
M_4	4	180
M ₅	8	180
M_6	4	180
M ₇	4	180
M ₈	0.24	180
M ₉	0.24	180
M ₁₀	0.48	180
M ₁₁	1	180

TABLE II. <u>SIZING OF TRANSISTOR</u>

B. Testing circuitry of SRAM:

Vddt sensors for fault free SRAM, under test SRAM, testing circuit, Operational amplifier and comparator shown in figure 5. Here under test SRAM compare to good SRAM by comparing by transient voltage of SRAM. Transient voltage is made by transient current converted from a Vddt sensor circuit. One block represented Vddt sensor circuit for good SRAM, which gives the transient voltage of good SRAM. Similarly Vddt sensor circuit for under test SRAM is made, which give another transient voltage of under test SRAM. The transient voltages of both SRAM are compared by comparator. In this project static comparator is used. Static comparator does not require any extra clock pulses. The output of comparator is depending on under test SRAM.



(An ISO 3297: 2007 Certified Organization)

Vol. 4, Issue 7, July 2015



Figure 5: BIST circuitry

C. Test for faulty SRAM array:

A Block level representation of the entire memory system [11] is shown in figure 6. The BIST architecture in the in same row share same word line. The 8×8 BIST architecture for SRAM is made by 64 bit BIST architecture. Here two decoders are used, which are column decoder and row decoder. Column decoder selects the bl and bl_bar and row decoder is used to select the word line (wl). The block level representation of memory system with testing circuit is shown in figure 6. The output of the particular BIST architecture gives the constant output or pulse output. If output is constant, i.e. the SRAM is non faulty otherwise SRAM is faulty.



Figure 6: 8×8 SRAM BIST Architecture



(An ISO 3297: 2007 Certified Organization)

Vol. 4, Issue 7, July 2015

IV.SIMULATION RESULT

A. Sensor output:

During the write operation, changes of state 0 to 1 or 1 to 0 in SRAM cell will cause the transient current to flow in SRAM cell. This transient current flows in good SRAM cell, which is shown in figure 7.



Figure 7: Transient current for good SRAM cell

This transient current is also flow in faulty SRAM cell flow in faulty SRAM cell, which is different from good SRAM cell. This is shown in figure 8.



PMOS M_1 senses the transient current. Due to these transient current, small spikes occurs at V (SENSE). This V (SENSE) voltage given to differential amplifier and output of differential amplifier is amplified by the amplifier. If SRAM cell is non faulty, then output of Vddt sensor circuit shown in figure 9.



Figure 9: Vddt sensor output voltage for good SRAM cell

If SRAM cell is non faulty, then output of Vddt sensor circuit shown in figure 10:



(An ISO 3297: 2007 Certified Organization)

Vol. 4, Issue 7, July 2015



Figure 10: Vddt sensor output voltage for faulty SRAM cell

B. Test for with single fault:

The test is performed with single fault, which is introduce by different value of resistance, then open and bridging faults are introduce individually. The difference in transient current between a faulty and fault free SRAM cell are shown in table II.

Resistances (R)Transient current for fault free SRAM cell (μA) Transient current for fault SRAM cell (μA) R_1 -33.747 -34.553 R_2 -33.747 -17.0698 R_2 -33.747 -39.905	
(R) fault free SRAM cell (μ A) fault SRAM cell (μ A) R1 -33.747 -34.553 R2 -33.747 -17.0698 R4 -33.747 -39.905	or
$\begin{array}{c ccccccccccccccccccccccccccccccccccc$	A)
$\begin{array}{c ccccccccccccccccccccccccccccccccccc$	
R_1 -33.747 -34.553 R_2 -33.747 -17.0698 R_2 -33.747 -39.905	
$\begin{array}{c c c c c c c c c c c c c c c c c c c $	
Ro _33.7/7 _39.905	
N3 -33.171 -37.703	
R ₄ -33.747 -5.52	
R ₅ -33.747 -23.47	
R ₆ -33.747 -22.701	
R ₇ -33.747 -40.367	
R ₈ -33.747 -22.701	
R ₉ -33.747 -25.537	
R ₁₀ -33.747 -39.67	
R ₁₁ -33.747 -22.699	

 TABLE II.

 TRANSIENT CURRENT FOR SINGLE FAULT

C. Output of BIST architecture:

The BIST architecture's output is shown in figure 11 & Figure 12. The comparator compares the output of Vddt circuit for good SRAM cell with Vddt circuit for under test SRAM cell. If under test SRAM cell is non faulty, then output of BIST architecture is a constant output.



(An ISO 3297: 2007 Certified Organization)

Vol. 4, Issue 7, July 2015



Figure 11: Output of BIST circuit for under test non faulty SRAM cell

If under test SRAM cell is faulty, then output of BIST architecture is a pulses output. Figure 12 shows the output put of BIST architecture with comparing to good SRAM cell and faulty SRAM cell.



Figure 12. Output of BIST circuit for under test faulty SRAM cell

V. CONCLUSION

In this research a BIST architecture using transient current method approach for fault detection is implemented and its effectiveness has been tested simple memory architecture with single or multiple faults, and also testing to large array of memory BIST architectures.

In perspective of future work on this project the consumption of BIST of SRAM for observing fault for such large array of memory, can be reduced by implementing a separate block for collective testing methodology for each block at once.

REFERENCES

- "International Technology Roadmap for semiconductors," Semiconductor Industry Association, 2003.
 "International Technology Roadmap for semiconductors," Semiconductor Industry Association, 2005.
- [3]Van de Goor, "Using March Test to test SRAMs," IEEE Design and Test of computers, pp. 8-14, March 1993.
- [4] Luigi Dilillo, Patrick Girard, "Dynamic Read Destructive in embedded-SRAM: Analysis and March Test Solution,"
- 9th IEEE European Test Symposium Congress Centre, Ajaccio, Corsica, France, pp. 132-137.
- [5] Rubio, J. Figueras, J. Segura, "Quiescent current sensor circuits in digital VLSI CMOS testing," Electronic Letters, Vol. 26, pp. 1204-120, 1990. [6] Bhavya Daya, Shu Jiang, Piotr Nowak, Jaffer Sharief, "Synchronous 16×8 SRAM Design".



(An ISO 3297: 2007 Certified Organization)

Vol. 4, Issue 7, July 2015

- [7] Sung-Mo Kang, Yusuf Leblebici, "CMOS Digital Integrated Circuits Analysis and Design," Third Edition, pp. 438-445.
- [8] Anumol K A, N.M.Siva Mangai, P.Karthigai Kumar, "Built in Self Test Architecture for Testing SRAM Using Transient Current Testing," IEEE Conference on Information and Communication Technologies, pp. 331-335, 2013.
- [9] Suriya A. Kumar, Rafic Z. Makki, and David Binkley, "IDDT Testing of CMOS Embedded SRAMs," Design, Automation and Test in Europe Conference and Exhibition, Proceedings, 2002.
- [10] Doe Yoon, Hong-Sik Kim, and Sungho Kang, "Dynamic Power Supply Current Testing for Open Defects in CMOS SRAMs," ETRI Journal, Volume 23, Number 2, pp., 77-84, June 2001.
- [11] Kiyoo Itoh, "VLSI memory chip design," Volume 5 of Springer Series in Advanced Microelectronics, 2010.