



An Improved DTC Scheme Using Fuzzy Logic for FSTPI-Fed Induction Motor Emulating the SSTPI Operation

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ABSTRACT: Direct Torque Control (DTC) of induction motor drives has become very popular and widely used in industrial applications due to its fast and good dynamic torque response. Induction motors (IM) are simple in construction and are less sensitive to the motor parameters compared to other vector control methods. The conventional DTC is based on flux and torque hysteresis controllers. Induction motor is fed from a Four Switch Three Phase Inverter (FSTPI) generating the voltage vectors of the Six Switch Three Phase Inverter (SSTPI) by emulation. Applying the most optimized voltage vector that produces the largest tangential component to the circular flux locus can accomplish the fastest dynamic torque response during transient states. The dynamic torque performance is very important in traction and electric vehicle application. A method to achieve fastest dynamic performance by modifying the two leg inverter fed DTC of induction motor based on Fuzzy Logic Concept (FLC) is discussed in this paper. FLC is one of the Artificial Intelligence methods that have found high application in most of the nonlinear systems like the electric motor drives. FLC can be used as a controller for any system without requirements of the system mathematical model unlike that of the conventional electric drive control, which uses the mathematical model. This paper presents a rule-based fuzzy logic controller scheme designed and applied for the speed control of an induction motor fed from a four switch three phase inverter emulating the six switch three phase inverter. Due to the usage of the FLC concept, the efficiency, reliability and performance of ac drive increases. Initial torque peak and torque ripple are minimized in the four switch three phase inverter based DTC using Fuzzy Logic.

KEYWORDS: Direct Torque Control (DTC), Four/Six Switch Three Phase Inverter (FSTPI/SSTPI), Fuzzy Logic Concept (FLC), Induction motor (IM), Emulation, Voltage Vectors

I. INTRODUCTION

DC motors were used extensively for variable speed drives applications in past decades due to the advantage that their flux and torque could be independently controlled by the field and armature current. Absence of mechanical commutator and brush assembly, rugged construction, simple structure and less maintenance requirement makes AC induction motors more acceptable than DC motors. In normal reference frames, the torque and flux of an induction motor are coupled together and hence independent control becomes difficult. Rapid advancement in high-power semiconductor devices and high speed digital signal processing has enabled decoupling of torque and flux giving rise to high performance induction motor drives capable of variable-speed performance. Now a day's induction motors are the work horse of the industry. The two important control methods of induction machines can be classified into scalar and vector control. Both Vector Control and Scalar control aim to control effectively the motor torque and flux in order to force the motor to accurately track the command trajectory regardless of machine and load parameter variations. In scalar control, the magnitude and frequency of voltage is controlled, whereas in vector control, the instantaneous position as well as the magnitude and frequency of voltage, current and flux linkage space vectors are controlled. Field Oriented Control (FOC) and Direct Torque Control (DTC) are the two most popular methods of vector control for high performance AC drives. Compared to FOC, DTC can provide extremely high dynamic response with very simple structure. The first DTC strategy involves a simple control scheme which makes it possible for rapid real time implementation as in [1]. The variable switching frequency and the ripples in flux and torque are the main drawbacks of conventional DTC. Hence several investigations are carried out in areas of high torque ripple and uncontrolled switching frequency which results in the development of several DTC strategies that have been proposed so far as in [3]-[4]. One of them is considering variable

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hysteresis band controllers as in [5]. Reconfiguration of SSTPI into a four-switch three inverter(FSTPI), in case of a switch/leg failure can be done where reliability is more important as in [6]-[7]. This paper proposes a new FLC based DTC strategy dedicated to FSTPI-fed IM drives. Fuzzy logic improves the performance of the FSTPI-fed DTC strategy of induction motor. It is based on the emulation of the SSTPI operation and the synthesis of an appropriate vector selection table, which is based on the hysteresis controllers and flux vector position as in [2]

II. FUZZY LOGIC BASED DTC OF FSTPI-FED INDUCTION MOTOR DRIVES

Direct Torque Control is a control technique used in AC drive system to obtain high performance torque control by which it is possible to control directly the stator flux and the torque by selecting the appropriate inverter switching state. Since its introduction, DTC based induction motors are being increasingly used in various industrial applications. The Direct torque control method is characterized by its simple implementation and a fast dynamic response.

A. Basic DTC

The conventional DTC drive using hysteresis comparators as two level flux controller and three level torque controller and their outputs are flux error and torque error respectively. It consists of flux and torque estimator, six switch three phase inverter and a voltage vector selection table. The SSTPI can generate eight balanced voltage vectors. Eight voltage vectors consists of six active voltage vectors and two zero voltage vectors. The active voltage vectors have the same magnitude and are phase shifted each by an angle of 60° . Direct Torque Control allow a direct control of the motor variables through an appropriate selection of the inverter control signals, in order to fulfill the requirements as whether the stator flux and torque need to be increased, decreased or maintained. These decisions are achieved according to the output C_ϕ of the flux hysteresis controller, the output C_τ of the torque hysteresis controller, and the angular displacement θ_s of the stator flux vector ϕ_s in the Clarke ($\alpha \beta$) plane. The torque and flux are controlled simultaneously by applying suitable voltage vectors of the eight balanced voltage vectors of the SSTPI.

B. FSTPI-fed DTC Strategy

The implementation scheme of the DTC strategy dedicated to the FSTPI-fed IM has the same layout as the one of the basic DTC strategy except that the SSTPI inverter is reconfigured to a FSTPI by removing one leg of the 3 leg inverter and the three level hysteresis controller in the torque loop is replaced by a two level hysteresis controller. This replacement is done because of the fact that no zero voltage vector is involved in the new DTC scheme.

1) *Intrinsic Voltage Vectors of FSTPI*: Two among the three phase terminals of the motor are connected to the FSTPI legs, while the third one is connected to the middle point of the dc-bus voltage. The voltage vectors of the FSTPI have unbalanced amplitudes and are phase shifted by an angle of $\frac{\pi}{2}$. Indeed, the vectors V_1 and V_3 have the same amplitude with magnitude of $\frac{V_{dc}}{\sqrt{6}}$ while vectors V_2 and V_4 have the same amplitude with a magnitude of $\frac{V_{dc}}{\sqrt{2}}$. The balanced voltage vectors of the SSTPI and unbalanced voltage vectors of FSTPI and corresponding switching states of the inverter switches are shown in Fig. 1.

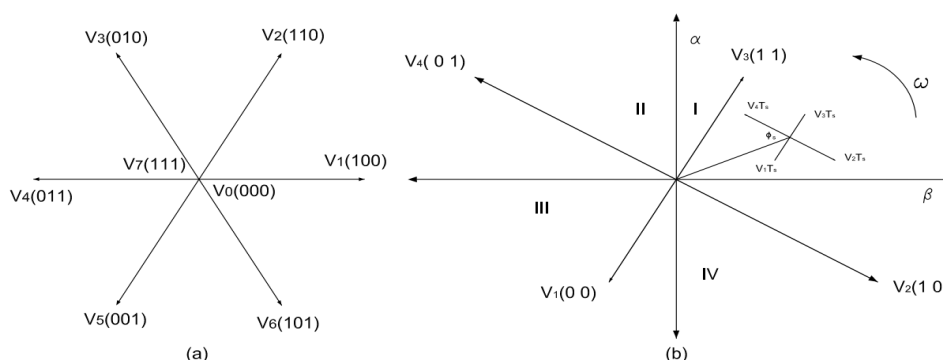


Fig.1 Voltage vectors and corresponding switching states of (a) SSTPI and (b) FSTPI

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2) *Generation of Balanced Voltage Vectors from FSTPI Voltage Vectors:* The six balanced voltage vectors of the SSTPI are generated using four unbalanced voltage vectors of the FSTPI. Thus emulates the whole SSTPI voltage vectors by switching the appropriate four FSTPI voltage vectors. The generated voltage vectors have the same amplitude and angular shift as those of the SSTPI. Based on the relationship between the balanced voltage vectors of the SSTPI and unbalanced voltage vectors of the FSTPI, the active voltage vectors V_k is the resultant of the any adjacent voltage vectors of the four voltage vectors of the FSTPI. The generated balanced six active voltage vectors $V_{11}, V_{12}, V_{23}, V_{33}, V_{34}$ and V_{41} are illustrated in the Fig. 2.

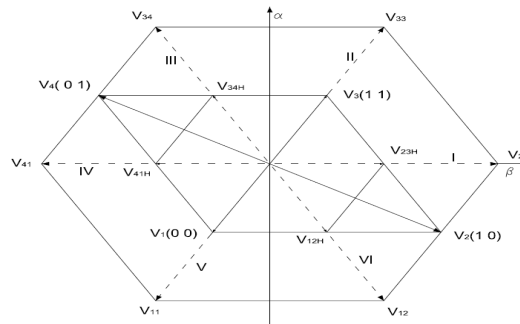


Fig. 2 Generation of Active Voltage Vectors Using the Four Unbalanced Voltage Vectors of the FSTPI

3) *Vector Selection Table of the New Strategy:* The three level torque controller in the conventional DTC is replaced by two level torque controller thereby reduce the switching frequency as well as the torque ripple. Each voltage vector of the SSTPI is obtained by applying suitable combination of the unbalanced voltage vectors of the FSTPI. In order to emulate the operation of the SSTPI, each control combination ($c \phi, c \tau$) should be maintained during two sampling periods $2T_s$. Implemented voltage vector selection table is shown in Table I. It is to be noted that both intrinsic and compounded voltage vectors are involved in sectors I, III, IV, and VI, while the sectors II and V, only the compounded voltage vectors are applied. Hence there may be an increase of the switching frequency in sectors II and V, with respect to the one in the remaining sectors.

TABLE I
IMPLEMENTED VOLTAGE VECTOR SELECTION TABLE

$c \phi$	+1		+1		-1		-1	
$c \tau$	+1		-1		+1		-1	
Periods T_s	1 st	2 nd	1 st	2 nd	1 st	2 nd	1 st	2 nd
Sector I	V_3		V_1	V_2	V_3	V_4	V_1	
Sector II	V_3	V_4	V_2	V_3	V_4	V_1	V_1	V_2
Sector III	V_4	V_1	V_3		V_1		V_2	V_3
Sector IV	V_1		V_3	V_4	V_1	V_2	V_3	
Sector V	V_1	V_2	V_4	V_1	V_2	V_3	V_3	V_4
Sector VI	V_2	V_3	V_1		V_3		V_4	V_1

C. Fuzzy Logic Based DTC

FLC is one of the Artificial Intelligence methods have found vast applications in most of the nonlinear systems like the electric motor drive. FLC can be used as controller for any system without requirements of the system mathematical model unlike that of the conventional electric drive control, which uses the mathematical model. Fuzzy Logic is an intelligent controller and its design do not depend on accurate mathematical model of the system and they can handle non-linearity of arbitrary intricacy. Among different intelligent algorithms fuzzy logic is the simplest which does not necessitate intensive mathematical analysis. Fuzzy Logic, unlike Boolean or crisp logic, deals with problems that have vagueness, uncertainty. Thus here the PI speed controller is replaced by fuzzy logic controller for getting better dynamic responses. In drive operation, the speed can be controlled indirectly by controlling the torque which, for the normal operating region, is directly proportional to the voltage to frequency ratio.

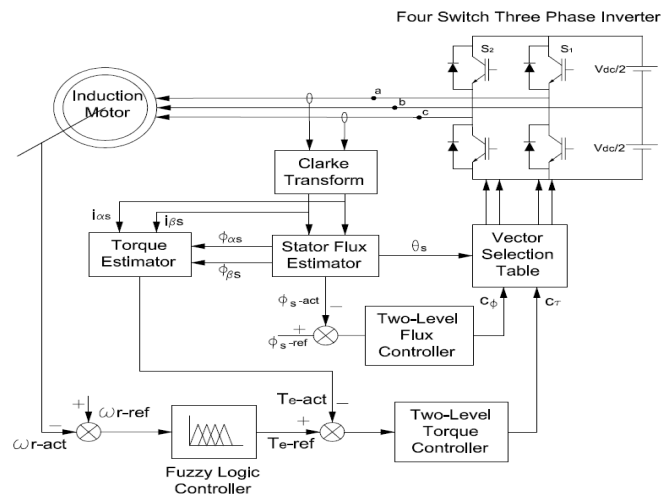


Fig. 3 Schematic Block Diagram of Fuzzy Logic Based FSTPI-fed DTC strategy

The schematic diagram of fuzzy logic based FSTPI-fed DTC strategy of induction motor is shown in Fig. 3. Usually the speed is controlled by PI controller and here PI controller is replaced by fuzzy logic controller whose output is the torque reference. The fuzzy logic controller is proposed to control the speed of the motor to be constant when the load varies. The speed error E and change of speed error CE are processed through the fuzzy logic controller whose output is the torque reference and it compared with the actual torque for the production of torque error. Triangular Membership functions were constructed to represent the input and output variables as shown in Fig. 4.

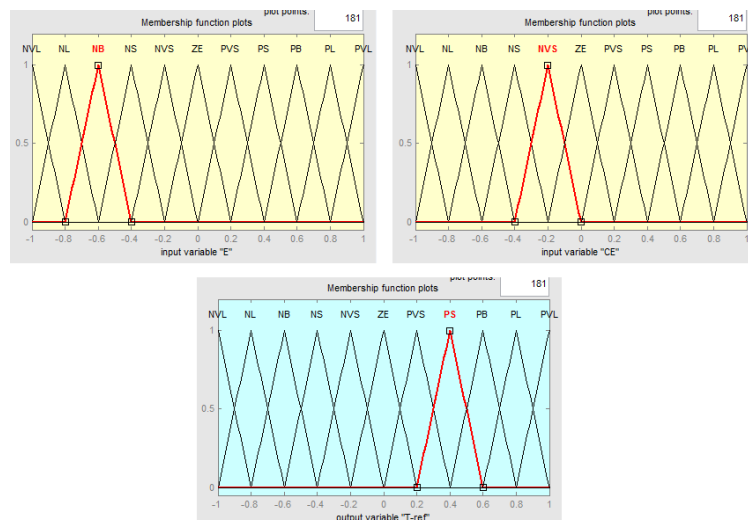


Fig. 4 Triangular membership functions of input and output variables

The speed error and the change in speed error are processed through the fuzzy logic controller whose output is the torque command. Consider the fuzzy speed control system, where the input signals are E and CE and the output is torque reference are fuzzified by assigning corresponding membership functions to each signal. The fuzzy sets are as follows: Z = Zero, PVS = Positive Very Small, NVS = Negative Very Small, PS = Positive Small, NS = Negative Small, PB = Positive Big, NB = Negative Big, PL = Positive Large, NL = Negative Large, PVL = Positive Very Large, NVL = Negative Very Large. The universe of discourse of all the variables, covering the whole region, is expressed in per unit values. The universe of discourse of inputs and outputs of the FLC are chosen between -1 and 1. There are



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eleven membership functions for E and CE whereas eleven membership functions for the output. Hence there may be $11 \times 11 = 121$ possible rules in the matrix. Table II shows the corresponding rule table for the speed controller. The top row and left column of the matrix indicate the fuzzy sets of the variables E and CE respectively and the output of the fuzzy controller is shown in the body of the matrix.

TABLE II
RULE BASE OF THE FUZZY SPEED CONTROLLER

CE\E	NVL	NL	NB	NS	NVS	ZE	PVS	PS	PB	PL	PVL
NVL	NVL	NVL	NL	NL	NB	NB	NS	NS	NVS	NVS	ZE
NL	NVL	NL	NL	NB	NB	NS	NS	NVS	NVS	ZE	PVS
NB	NL	NL	NB	NB	NS	NS	NVS	NVS	ZE	PVS	PVS
NS	NL	NB	NB	NS	NS	NVS	NVS	ZE	PVS	PVS	PS
NVS	NB	NB	NS	NS	NVS	NVS	ZE	PVS	PVS	PS	PS
ZE	NB	NS	NS	NVS	NVS	ZE	PVS	PVS	PS	PS	PB
PVS	NS	NS	NVS	NVS	ZE	PVS	PVS	PS	PS	PB	PB
PS	NS	NVS	NVS	ZE	PVS	PVS	PS	PS	PB	PB	PL
PB	NVS	NVS	ZE	PVS	PVS	PS	PS	PB	PB	PL	PL
PL	NVS	ZE	PVS	PVS	PS	PS	PB	PB	PL	PL	PVL
PVL	ZE	PVS	PVS	PS	PS	PB	PB	PL	PL	PVL	PVL

III. SIMULATION STUDIES AND COMPARISON

Simulations of basic DTC, FSTPI-fed DTC and fuzzy logic based FSTPI-fed DTC are carried out for evaluating the performance of each DTC scheme using MATLAB Simulink. There are many benefits of simulation in the process of analysis, design and evaluation of power converters. The Induction motor parameters and ratings are shown in Table III.

TABLE III
INDUCTION MOTOR PARAMETERS AND RATINGS

Parameter	Value	Rating	Value
No of pole pairs	2	Rated Shaft Power	0.75kW
R_s	11.765 Ω		
R_r	6.112 Ω	Line to Lin Voltage	415V
L_s	0.666H		
L_r	0.666H	Rated Current	1.8 A
L_m	0.6325H	Rated Speed	1400rpm
J	0.001Kgm ²	Frequency	50Hz

The band width of the stator flux controller and electromagnetic torque controller is equal to ± 0.02 Wb and ± 0.04 Nm respectively. The dc bus voltage is limited to 400V for the sake of safe operation of the inverter. Normally one goes about the simulation procedure by bringing in fundamental blocks from the library and suitably interconnecting them to perform the task of a subsystem. It is possible to use this subsystem obtained by grouping these fundamental blocks as a single block wherever such a subsystem is to be connected. Various subsystem blocks so generated are interconnected to simulate the entire system. The amplitude of equivalent voltage vectors generated by FSTPI is the half of the one of the voltage vectors generated by SSTPI.

A. FLC based FSTPI-fed DTC

The speed and torque responses at steady state operation of the IM under the control of the basic DTC strategy of induction motor, for a speed of 300rpm and a constant load torque $T_l = 5$ Nm is obtained from Simulation in MATLAB Simulink which is shown in Fig. 5.

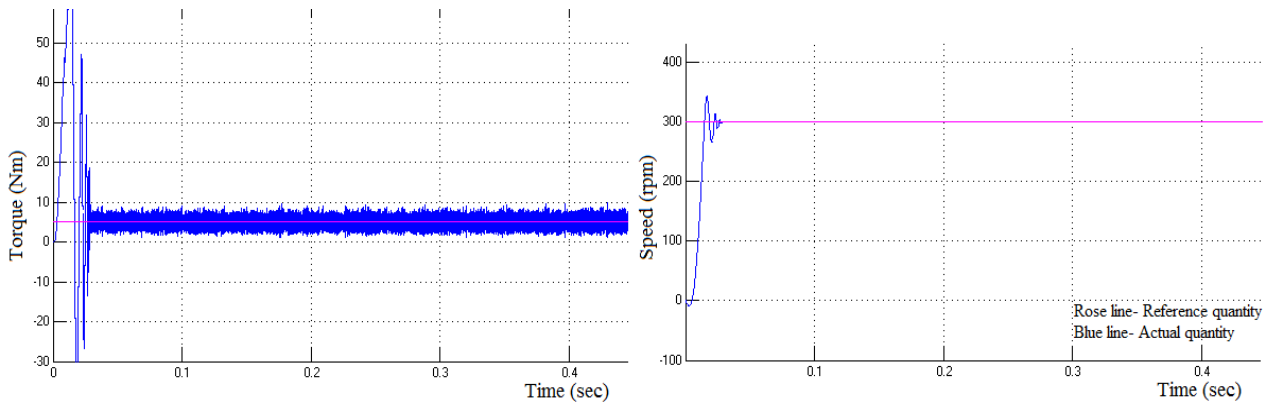


Fig. 5 Speed and torque responses of the basic DTC of IM drives

B. FSTPI-fed DTC Strategy

Simulation and analysis of the FSTPI-fed induction motor is done in MATLAB Simulink. The speed and torque response characterizing the steady state operation of the IM under the control of the FSTPI-fed DTC of IM drive for a speed of 300rpm and a constant load torque $T_1=5$ Nm are shown in Fig. 6. The initial transients of torque and torque ripple of FSTPI-fed DTC are reduced in some extent than that of the basic DTC of IM drives

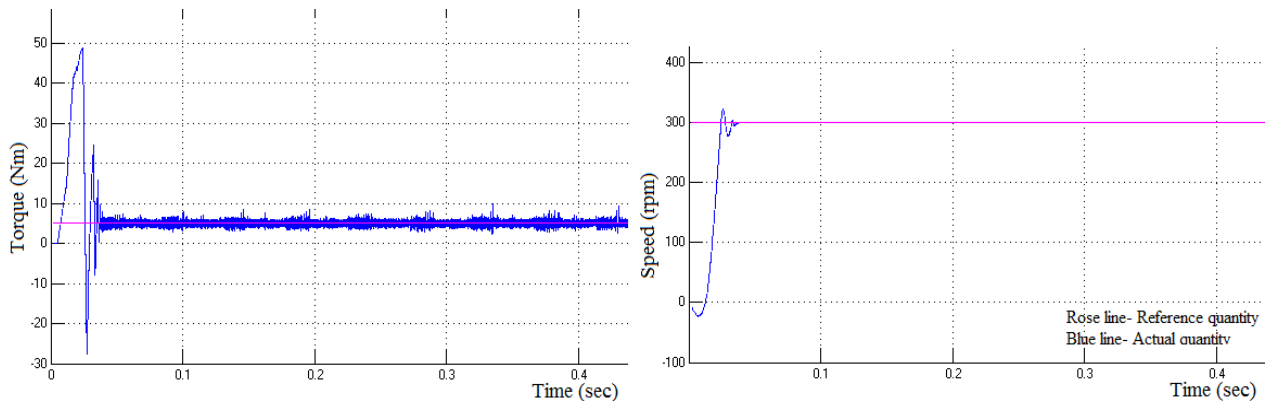


Fig. 6 Speed and torque response of FSTPI-fed DTC of IM drives

C. FLC based FSTPI-fed DTC

Simulation of fuzzy logic based FSTPI-fed DTC of induction motor is done in MATLAB Simulink and analysed waveforms. The speed and torque responses characterizing the steady state operation of the IM under the control of the proposed DTC strategy, for a speed of 300rpm and a constant load torque $T_1=5$ Nm are shown in Fig. 9. The torque response curve of each DTC strategy shows that torque ripples and the initial hike of the torque curve are reduced in considerable amount in the strategy having fuzzy logic controller. The fuzzy logic controller block and FSTPI-fed fuzzy logic based DTC of IM in MATLAB Simulink are shown in Fig.7 and Fig. 8 respectively.

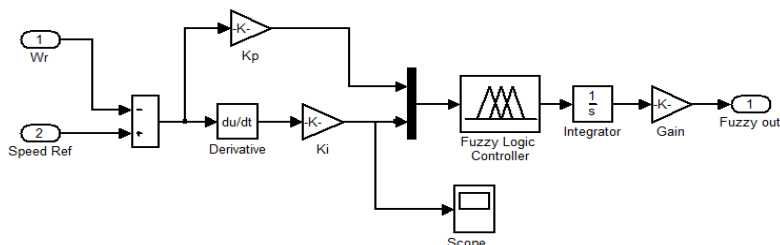


Fig. 7 MATLAB Simulink model of fuzzy logic controller

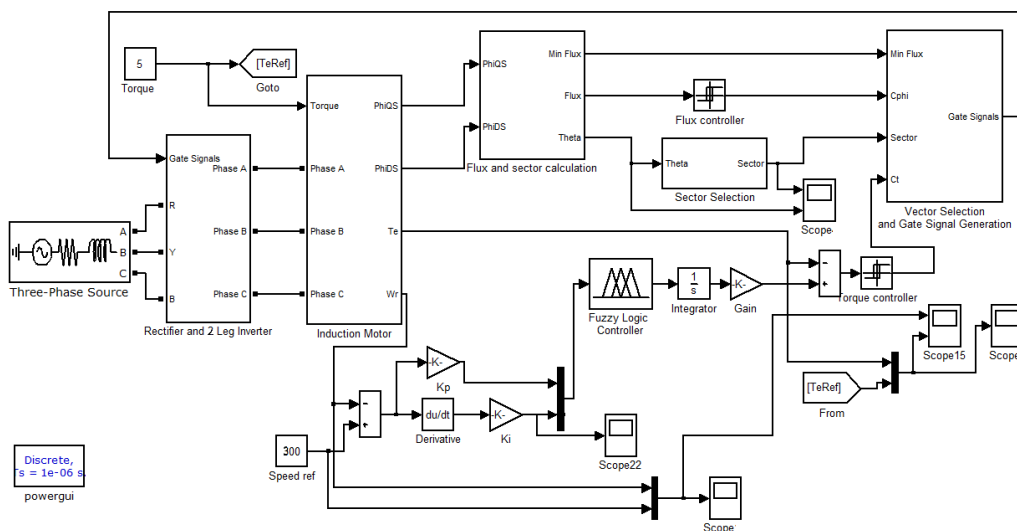


Fig. 8 MATLAB Simulink model of FSTPI-fed fuzzy logic based DTC of IM drives

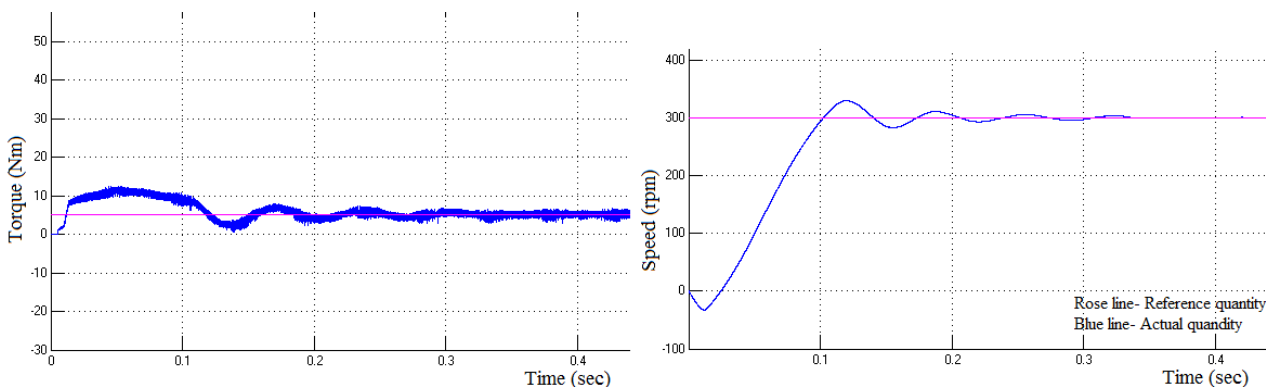


Fig. 9 Speed and torque responses of fuzzy logic based FSTPI-fed DTC strategy of IM drive

IV. CONCLUSION

Direct Torque Control is a dominating control technique for induction motor among several control techniques. Its principles and basic concepts have been introduced and thoroughly explained. It is clear from the explanations that the



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method of direct torque control also allows the decoupled control of motor torque and motor stator flux. According to the analysis, the DTC strategy is simpler to implement than the flux vector control method because it does not require voltage modulators and co-ordinate transformations and it reduces undesired torque ripple. This thesis mainly describes the analysis of conventional DTC, DTC of FSTPI fed induction motor and fuzzy logic based DTC. Simulations of various types of DTC are done in MATLAB by emulating the voltage vectors of SSTPI from the voltage vectors of FSTPI. Fuzzy logic based controller is used to reduce the torque ripple and it was reduced in considerable percentage. Simulation results show the validity of the fuzzy logic based DTC and it achieves a considerable reduction in torque and flux ripple signal as well as initial hike of torque almost avoided.

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REFERENCES

- [1] Takahashi, Isao, and Toshihiko Noguchi. "A new quick-response and high-efficiency control strategy of an induction motor." *Industry Applications, IEEE Transactions on* 5 (1986): 820-827.
- [2] El Badsı, Bassem, Badi Bouzidi, and Ahmed Masmoudi. "DTC scheme for a four-switch inverter-fed induction motor emulating the six-switch inverter operation." *Power Electronics, IEEE Transactions on* 28.7 (2013): 3528-3538.
- [3] Zhang, Yongchang, and Jianguo Zhu. "Direct torque control of permanent magnet synchronous motor with reduced torque ripple and commutation frequency." *Power Electronics, IEEE Transactions on* 26.1 (2011): 235-248
- [4] Taheri, Asghar, Abdolreza Rahmati, and Shahriyar Kaboli. "Efficiency improvement in DTC of Six-phase induction machine by adaptive gradient descent of flux." *Power Electronics, IEEE Transactions on* 27.3 (2012): 1552-1562.
- [5] Kang, Jun-Koo, Dae-Woong Chung, and Seung-Ki Sul. "Direct torque control of induction machine with variable amplitude control of flux and torque hysteresis bands." *Electric Machines and Drives, 1999. International Conference IEMD'99*. IEEE, 1999.
- [6] Beltrao de Rossiter Correa, Mauricio, et al. "A general PWM strategy for four-switch three-phase inverters." *Power Electronics, IEEE Transactions on* 21.6 (2006): 1618-1627.
- [7] Wang, Rui, Jin Zhao, and Yang Liu. "A comprehensive investigation of four-switch three-phase voltage source inverter based on double Fourier integral analysis." *Power Electronics, IEEE Transactions on* 26.10 (2011): 2774-2787
- [8] Zhang, Yongchang, et al. "An improved direct torque control for three-level inverter-fed induction motor sensorless drive." *Power Electronics, IEEE Transactions on* 27.3 (2012): 1502-1513.
- [9] Bose, Bimal K. *Modern power electronics and AC drives*. Vol. 123. USA: Prentice Hall, 2002.

BIOGRAPHY



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