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ijareeie@gmail.com

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# Harmonic Minimization for Cascade Multi Level inverter based on Differential Evolutionary Algorithm

R.Gunasekaran<sup>1</sup>, Dr.V.S.Arulmurugan<sup>2</sup>, C.Shalini<sup>3</sup>, B.Sonaa<sup>3</sup>, G.Saravanan<sup>3</sup>, S.Soundhar Raj<sup>3</sup>

Assistant Professor, Department of EEE, Excel College of Engineering and Technology, Komarapalayam, Tamilnadu, India <sup>1</sup>

Professor, Department of EEE, Excel College of Engineering and Technology, Komarapalayam, Tamilnadu, India <sup>2</sup>

UG Student, Department of EEE, Excel College of Engineering and Technology, Komarapalayam, Tamilnadu, India <sup>3</sup>

**ABSTRACT:** A new topology of cascaded multilevel converter is proposed. The proposed to is predicated on a cascaded connection of single-phase sub multilevel converter units and full bridge converters. Compared to standard multilevel converter, the amount of DC voltage sources, switches, installation area, and converter cost is reduced as number of voltage steps increases. structure of proposed topology is optimized so as to utilize a minimum number of switches and DC voltage sources, produce a high number of output voltage steps. The H-bridge based multilevel inverter increase the amount of output voltage levels by adding switch components and DC input voltage sources. lower number of conducting switches which reduces the switching losses. Comparison study is presented between the proposed inverter and other recently presented to apologies

## I. INTRODUCTION

Multilevel inverters are more advanced and latest type of power electronic converters that synthesize a desired output voltage from several levels of dc voltages as inputs. By taking sufficient number of dc sources, anearly sinusoidal voltage waveform are often synthesized.

In comparison with the hard-switched two-level pulsewidth modulation inverters, multilevel inverters offerseveral advantages including their capabilities to workat high voltage with lower voltage stress per switching,high efficiency and low electromagnetic interferences [1],[3] etc.To synthesize multilevel output ac voltage usingdifferent levels of dc inputs, semiconductor devices must

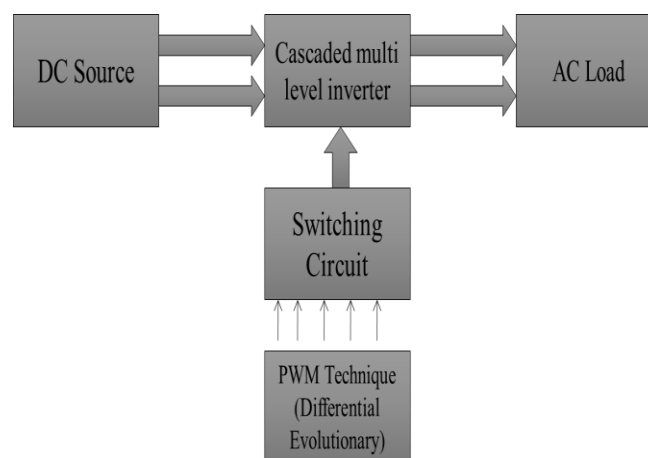
be switched on and off in such how that desiredfundamental is obtained with minimumharmonicdistortion. There are different approaches for theselection of switching techniques for the multilevelinverters [4]-[9], one among the important techniques isselective harmonic elimination (SHE) method. In SHEtechnique, certain predominating lower order harmonicsare eliminated whereas higher order harmonics arefiltered using suitable filter. Switching angles arecomputed by solving the SHE equations, but it's difficultto solve SHE equations due to their nonlinearcharacteristics. thanks to nonlinear nature, solution ofthoseequations could also be simple, multiple or maybe no solution fora particular value of modulation index (m). an enormous task ishow to get all possible solutionsonce they exist using simple and fewer computationally complex method. Oncethese solutions are obtained, the solutions having leastTHD are selected for switching purpose.In [4], [5], iterativenumerical techniques likeNewton- Raphsonmethod are implemented tosolve the SHE equations producing just one solution set,and even for this a correct initial guess and starting valueof m that solutions exist are required. Sometechniques as discussed in [6], [7], here SHE equationsare first converted intopolynomial equations, and then the resulting polynomial equations are solved usingtheory of resultants of polynomials and therefore the theory ofsymmetrical polynomials, producing all possiblesolutions. an issue with these approaches is that forhigher levels of multilevel inverter the order ofpolynomials becomes very high, thereby making thecomputations of solutions of those polynomial equationsvery complex. Optimization techniques supported GeneticAlgorithm(GA)and Differential evolutionary Algorithm are discussed in [8], [9] for computingswitching

anglesjustfor7levelmultilevelinverters.Theimplementationofthose approachrequiresproperselectionofcertainparameters like population size,mutation rate, initial weight etc. It becomes difficult toselect these parameters for higher level multilevelinverters. to bypass above



mentioned problems, a simple optimization technique supported sequential quadratic programming (SQP) is proposed during this paper to solve SHE equations which produces all possible solutions. The proposed technique is implemented in such a way that each one possible solution for any number of H-bridges connected serial are computed by using any arbitrary initial guess with negligible computational effort. An entire analysis for an 11-level inverter using five H-bridges per introduce series is presented, and it's shown that for a variety of  $m$ , switching angles are often computed to supply the specified fundamental voltage along with elimination of 5th, 7th, 11th, and 13th order harmonic components. The computational results are validated through experiments.

#### BLOCK DIAGRAM:



## II. CASCADE MULTILEVEL INVERTER

Cascade Multilevel Inverter (CMLI) is one among the foremost important topology within the family of multilevel inverters. It requires least number of components with compare to diode-clamped and flying capacitors type multilevel inverters. It's modular structure with simple switching strategy and occupies less space [1] - [3].

The CMLI consists of variety of H-bridge inverter units with separate dc source for every unit and is connected in cascade or series as shown in Fig. 1. Each H-bridge can produce three different voltage levels:  $+V_{dc}$ , 0, and  $-V_{dc}$  by connecting the dc source to ac output side by different combinations of 6 switches  $S_1, S_2, S_3, S_4, S_5$  and  $S_6$ . ac output of every H-bridge is connected serial such that the synthesized output voltage waveform is that the sum of all of the individual H-bridge outputs.

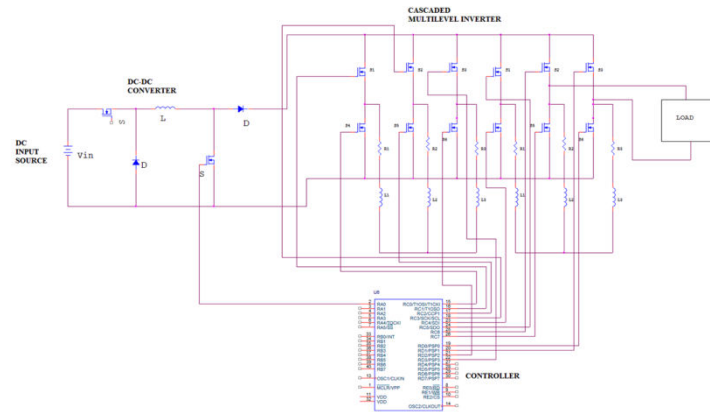
cascade and using proper modulation scheme, a nearly sinusoidal output voltage waveform are often synthesized. The number of levels within the output phase voltage is  $2s+1$ , where  $s$  is that the number of H-bridges used per phase. Fig. 2 shows an 11-level output phase voltage waveform using five H-bridges. The magnitude of the ac output phase voltage is given by

$$v_{an} = v_{a1} + v_{a2} + v_{a3} + v_{a4} + v_{a5} \quad [2].$$

In general, when  $s$  number of H-bridges per phase is connected in cascade, the Fourier series expansion of the staircase output voltage waveform is given by (1).



CIRCUIT DIAGRAM:



### III.HARMONIC ELIMINATION CONTROL TECHNIQUE USING EVOLUTIONARY ALGORITHMS USING GA & DEA

Harmonic Elimination pulse width modulation (HEPWM) method has been widely applied to remove harmonics due to its superior frequency spectra. It requires the answer of a set of transcendental nonlinear equations. Softcomputing(SC) methods are extensively employed to unravel this problem due to their effective global search ability. Genetic Algorithm (GA) and Differential evolution (DE)

has surpassed most of the SC methods in diverse fields but it has never been utilized to unravel this problem. In this work. GA and DE is employed to unravel the HEPWM problem for eleven level cascaded multilevel voltage source inverter (MCSI). Simulation results have shown that the discontinuities of the HEPWM angle trajectories are nullified and a wider over-modulation range has been covered, enhancing the use of DC link voltages and extending the appliance of HEPWM for top power applications.

### IV.FORMULATING THE MATTER

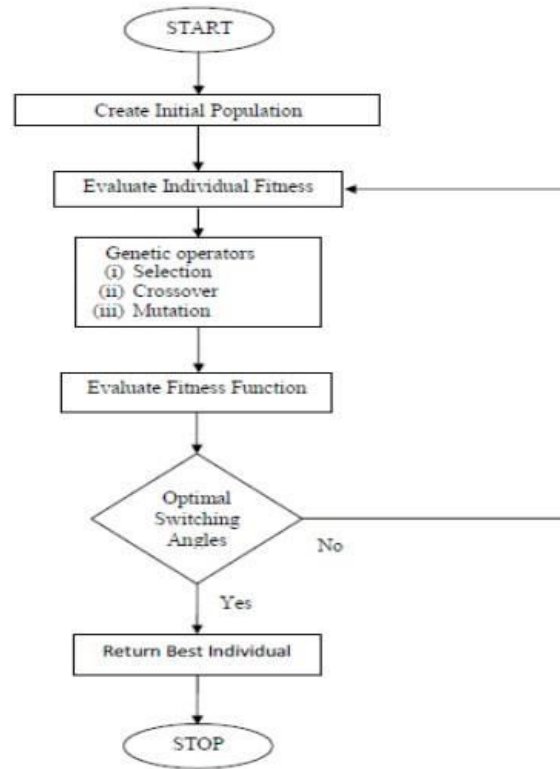
The step- by- step procedure to unravel the SHE problem with unequal dc sources using GA is as follows.

- i) Get the info for the system. At the primary step, the required parameters of the algorithm like population size, modulation index (M), Nominal Voltage, Number of Inverter level, max iteration number are determined.
- ii) Random population generation.
- iii) Fitness function – the fitness evaluation evaluates the population using the fitness function given by equation .
- iv) Parent Selection – Best parents of generation are selected supported the roulette. Wheel selection for creating next generation.
- v) Crossover – the crossover operator creates the two new child vector by mating the 2 best parents using arithmetic crossover method.
- vi) Mutation – the mutation operator mutates a toddler by changing any of its genes.
- vii) Survival Selection – the survival selection operator chooses the vectors that are getting to compose the population within the next generation.





**V. FLOW CHART OF GENETIC ALGORITHM**

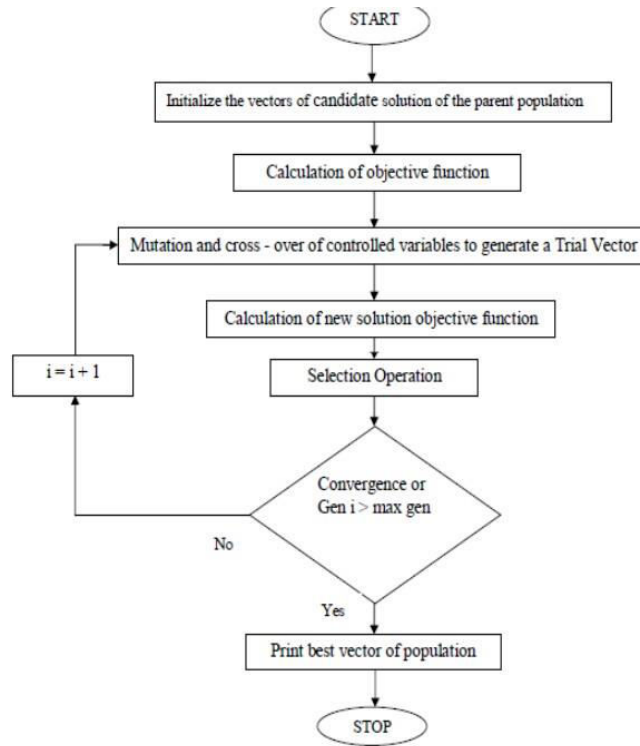


**VI. DIFFERENTIAL EVOLUTION ALGORITHM**

The DE algorithm may be a population based algorithm like genetic algorithms using the similar operators; crossover, mutation and selection. The most difference in constructing better solutions is that genetic algorithms on crossover while DE relies on mutation operation. This main operation is predicated on the differences of randomly sampled pairs of solutions within the population. The algorithm uses mutation operation as an enquiry mechanism and selection operation to direct the search toward the prospective regions within the search space. The Differential Evolutionary algorithm also uses a non-uniform crossover which will take child vector parameters from one parent more often than it does from others. The recombination (crossover) operator efficiently shuffles information about successful combinations, enabling the look for a far better solutionspace. An optimization task consisting of D parameters can be represented by a D-dimensional vector. In DE, a population of NP solution vectors is randomly created at the start. This population is successfully improved by applying mutation, crossover and selection operators.



**FLOW CHART DEA**



**VII. STEPS OF DIFFERENTIAL EVOLUTION**

The main steps of the DE algorithm are given below:

- Initialization
- Mutation
- Crossover
- Selection

**Mutation**

For each target vector  $x_{i,G}, i= 1,2,3,\dots, NP$ , a mutant vector is produced by

$$v_{i,G+1} = x_{r1,G} + F \cdot (x_{r2,G} - x_{r3,G})$$

with random indexes  $r1, r2, r3 \{1,2,\dots, NP\}$ , integer, mutually different and  $F > 0$ . The mutation factor  $F$  is a constant from  $[0,2]$  which controls the amplification of the differential variation

$$(x_{r2,G} - x_{r3,G})$$

**Crossover**

In order to increase the diversity of the parameter vectors, crossover is introduced. To this end, the trial vector:

$$u_{i,G+1} = (u1_{i,G+1}, u2_{i,G+1}, \dots, uDi_{i,G+1})$$

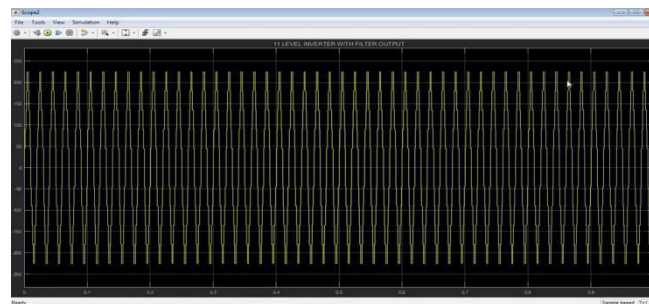
is formed, where



$$u_{ji,G+1} = \begin{cases} v_{ji,G+1} & \text{if } (rand_j \leq CR) \text{ or } j = I_{rand} \\ x_{ji,G} & \text{if } (rand_j > CR) \text{ and } j \neq I_{rand} \end{cases}$$

### Selection

To decide whether or not it should become a member of generation  $G+1$ , the trial vector  $u_{i,G+1}$  is compared to the target vector  $x_{i,G}$  using the greedy criterion. If vector  $u_{i,G+1}$  yields a smaller cost function value than  $x_{i,G}$ , then  $x_{i,G+1}$  is set to  $u_{i,G+1}$ ; otherwise, the old value  $x_{i,G}$  is retained.



Output wave form

### VIII. CONCLUSION

The selective harmonic elimination method at fundamental frequency switching scheme has been implemented using the optimization technique that produces all possible solution sets once they exist. In comparison with other suggested methods, the proposed technique has many advantages such as: it can produce all possible solution sets for any numbers of multilevel inverter without much computational burden, speed of convergence is fast, it can produce continuous solutions for the entire range of modulation index thereby giving more flexibility on top of things action etc. The proposed technique was successfully implemented for computing the switching angles for 11-level CMLI. An entire analysis for 11-level inverter has been presented and it's shown that a big amount of THD reduction can be attained if all possible solution sets are computed.

### ACKNOWLEDGMENT

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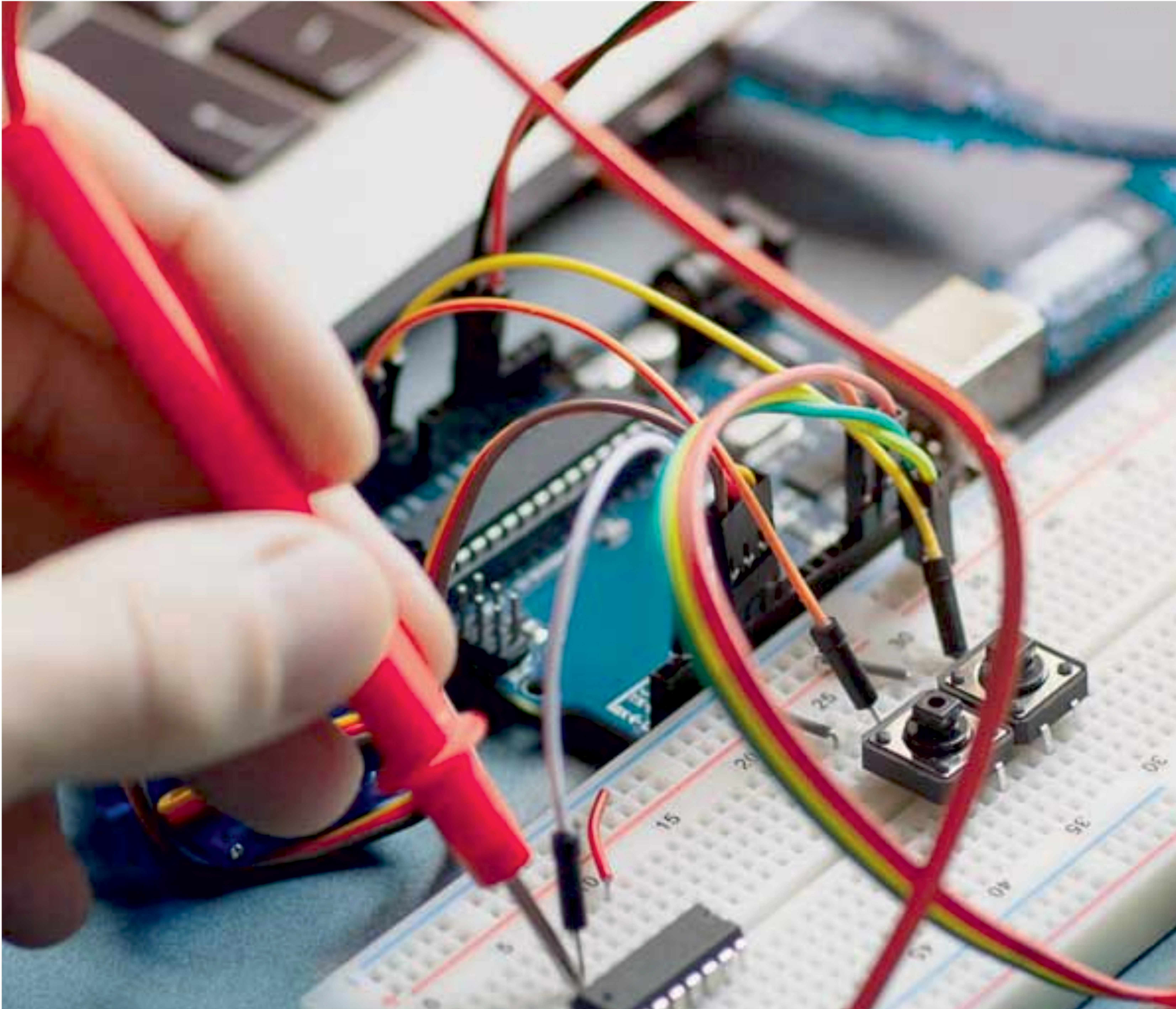
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