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# Stability Analysis of Distributed Generation System Connected to STATCOM Using Variable DC Voltage and Constant Modulation Index

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**ABSTRACT**: Power demand is increasing day by day, the new generation requires enough of resources, time period and budget allocation, and it is observed in the transmission system the percentage of power is wasted in the form of loss, and these is especially due to reactive power, by proper compensation of the reactive power through a FACTS device, known as statcom, the power can be effectively transmitted and can meet the demand. In this paper The STATCOM (Synchronous Static Compensator) connected with variable DC voltage source and constant modulation index is used to supply dynamic VARs required during power system faults for voltage support. All responses are obtained through MATLAB SIMULINK tool box

**KEYWORDS:**STATCOM, Distributed generation system, variable DC voltage STATCOM, power flicker.

### **I.INTRODUCTION**

Power Generation and Transmission is a complex process, requiring the working of many components of the power system in tandem to maximize the output. One of the main components to form a major part is the reactive power in the system. It is required to maintain the voltage to deliver the active power through the lines. Loads like motor loads and other loads require reactive power for their operation. To improve the performance of ac power systems, we need to manage this reactive power in an efficient way and this is known as reactive power compensation. There are two aspects to the problem of reactive power compensation: load compensation and voltage support. Load compensation consists of improvement in power factor, balancing of real power drawn from the supply, better voltage regulation, etc. of large fluctuating loads. Voltage support consists of reduction of voltage fluctuation at a given terminal of the transmission line. Two types of compensation can be used: series and shunt compensation. These modify the parameters of the system to give enhanced VAR compensation

In recent years, static VAR compensators like the STATCOM have been developed. These quite satisfactorily do the job of absorbing or generating reactive power with a faster time response and come under Flexible AC Transmission Systems (FACTS). This allows an increase in transfer of apparent power through a transmission line, and much better stability by the adjustment of parameters that govern the power system i.e. current, voltage, phase angle, frequency and impedance.

In this paper, first, a popularly accepted mathematical model is derived for the STATCOM. Then, the control strategy of the model which been proposed and described. The controller for this model has also been derived. This model has been simulated with designed controller by variation of pre-charge voltage on dc-link of the STATCOM.Finally simulation results are presented and demonstrated

A double generator system and a transmission line where an ideal shunt compensator is connected to the middle of the line. The generators have an equivalent reactance of XG1 and XG2 while the transmission line has an equivalent reactance of XdL. The voltages at Point of Common Coupling (PCC) of the generators are given as  $V1 \angle \delta1$  and  $V2 \angle \delta2$ .



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The shunt compensator connected in the middle of the line is a voltage source that is continuously controlled to VSC $\angle \delta$ SC.



Figure 1 Connection of an ideal shunt compensator to transmission line



Figure 2 Phasor diagram of shunt compensation. (a) Reactive power compensation, and (b) reactive and active power compensation.

The phase differences show that I1–SC current flows from first generator to the line while ISC–2 flows from line to the second generator. The ISC phasor is the derived current flowing through the shunt capacitor where it is orthogonal to the Vsc as seen in Fig. 2(a). This means the compensator does not exchange active power (P) with the line. In this case, the compensator has only reactive power on its connections. Hence, the power transferred from V1 to V2 can be calculated as

$$P_1 = \frac{2V^2}{X_L} \sin(\delta/2) \qquad ----(1)$$

Where the P1 is the active power supplied by V1 while V is the vector sum of V1 and V2 sources. In case of any compensator was not included in the system, the transferred power would be expressed as given in the following,

$$P_1 = \frac{V^2}{X_L} \sin(\delta) \qquad \qquad ----(2)$$

Where the compensator increases the power control ability of the transmission line since  $2\sin(\delta/2)$  yields higher value than  $\sin(\delta)$  in the range of  $[0, 2\pi]$ . In case of the phase angles of V1 and V2 are different from  $\delta/2$ , this situation causes the power flowing through sources has active and reactive power components as shown in Fig. 1.2b. In this situation, the shunt compensator owing to its power electronics based structure can be operated to adjust one of active or reactive power. Besides, the device structure also varies according to requirement of active or reactive power compensation



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since they are different in terms of energy storage elements. The connection type and switching device of compensators involve several different operating characteristics. The thyristor based topologies are classified into two main groups of self-commutated and force-commutated.

#### **II.SYSTEM MODEL AND ASSUMPTIONS**

#### A. Operating Principle

As is well known, the STATCOM is, in principle, a static (power electronic) replacement of the age-old synchronous condenser. Fig.4 shows the schematic diagram of the STATCOM at PCC through coupling inductors. The fundamental phasor diagram of the STATCOM terminal voltage with the voltage at PCC for an inductive load in operation. Ideally, increasing the amplitude of the STATCOM terminal voltage Voa above the amplitude of the utility voltage Vsa causes leading (capacitive) current Ica to be injected into the system at PC.

### **B.** Modelling

The modelling of the STATCOM, though well known, is reviewed in the lines below, for the sake of convenience. The modeling is carried out with the following assumptions:

- 1. All switches are ideal
- 2. The source voltages are balanced
- 3. Rs represents the converter losses and the losses of the

coupling inductor

4. The harmonic contents caused by switching action are negligible

The 3-phase stationary abc coordinate vectors with 1200 apart from each other are converted into  $\alpha\beta$  2-phase stationary coordinates (which are in quadrature). The  $\alpha$  axis is aligned with a axis and leading  $\beta$  axis and both converted into dq two-phase rotating coordinates. The Park's abc to dq transformation matrix is

$$K = \sqrt{\frac{2}{3}} \begin{bmatrix} \cos(\omega t) & \cos(\omega t - 2\pi/3) & \cos(\omega t + 2\pi/3) \\ \sin(\omega t) & \sin(\omega t - 2\pi/3) & \sin(\omega t + 2\pi/3) \\ 1/\sqrt{2} & 1/\sqrt{2} & 1/\sqrt{2} \end{bmatrix} - -(3)$$

The actual proposed circuit is too complex to analyze as a whole, so that it is partitioned into several basic sub-circuits. The 3-phase system voltage vs.abc lagging with the phase angle  $\alpha$  to the STATCOM output voltage vo,abc and differential form of the STATCOM currents are defined in (4) and (5).

$$v_{s,abc} = \begin{bmatrix} v_{sa} \\ v_{sb} \\ v_{sc} \end{bmatrix} = \sqrt{\frac{2}{3}} V_s \begin{bmatrix} \sin(\omega t - \alpha) \\ \sin(\omega t - \alpha - \frac{2\pi}{3}) \\ \sin(\omega t - \alpha + \frac{2\pi}{3}) \end{bmatrix} \qquad ----(4)$$
$$L_s \frac{d}{dt} (i_{c,abc}) = -R_s i_{c,abc} + v_{s,abc} - v_{0,abc} \qquad ----(5)$$

Where, Vs ,  $\omega$  , Rs , and Ls have their usual connotations. The above voltages and currents are transformed into dq frame

$$L_{s} \frac{d}{dt}(i_{cq}) = -R_{s}i_{cq} - wL_{s}i_{cd} + v_{sq} - v_{oq} - - - - (6)$$

$$L_{s} \frac{d}{dt}(i_{cd}) = wL_{s}i_{cq} - R_{s}i_{cd} + v_{sd} - v_{od} - - - - (7)$$

The switching function S of the STATCOM can be defined as follows  $\int \sin(\omega t) dt = 1$ 

$$S = \begin{bmatrix} S_a \\ S_b \\ S_c \end{bmatrix} = \sqrt{\frac{2}{3}m} \begin{bmatrix} \sin(\omega t) \\ \sin(\omega t - \frac{2\pi}{3}) \\ \sin(\omega t + \frac{2\pi}{3}) \end{bmatrix} - - - - (8)$$

The modulation index, being constant for a programmed PWM, is given by,



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The STATCOM output voltages in dq transformation are  $v_{o,qdo} = m[0 \ 1 \ 0]^T v_{dc} \qquad ----(10)$ The DC side current in the capacitor in dq transformation  $i_{dc} = m[0 \ 1 \ 0] [i_{cq} i_{cd} i_{co}]^T \qquad ----(11)$ The voltage and current related in the dc side is given by  $\frac{dv_{dc}}{dt} = \frac{m}{C} i_{cd} \qquad ----(12)$ 

The complete mathematical model of the STATCOM in dq frame is obtained as given in (13)

$$\frac{d}{dt} \begin{bmatrix} i_{cq} \\ i_{cd} \\ v_{dc} \end{bmatrix} = \begin{bmatrix} -\frac{R_s}{L_s} & -w & 0 \\ w & -\frac{R_s}{L_s} & -\frac{m}{L_s} \\ 0 & \frac{m}{c} & 0 \end{bmatrix} \begin{bmatrix} i_{cq} \\ i_{cd} \\ v_{dc} \end{bmatrix} + \frac{V_s}{L_s} \begin{bmatrix} -\sin\alpha \\ \cos\alpha \\ 0 \end{bmatrix} \quad --(13)$$

#### **Design of controller**

With the assumption of the system voltage and STATCOM output voltage are in phase and hence the equation (13) can be modified as given in equation (14)

$$\frac{d}{dt} \begin{bmatrix} i_{cq} \\ i_{cd} \end{bmatrix} = \begin{bmatrix} -\frac{R_s}{L_s} & -\omega \\ \omega & -\frac{R_s}{L_s} \end{bmatrix} \begin{bmatrix} i_{cq} \\ i_{cd} \end{bmatrix} + \frac{1}{L_s} \begin{bmatrix} v_{sq} \\ v_{sd} \end{bmatrix} - \begin{bmatrix} v_{oq} \\ v_{od} \end{bmatrix} \end{bmatrix} - -(14)$$

So the equation (14) is a Multiple Input and Multiple Output (MIMO) system and its input and output are given as  $[u] = \begin{bmatrix} v_{oq} \\ v_{od} \end{bmatrix}, [y] = \begin{bmatrix} i_{cq} \\ i_{cd} \end{bmatrix} \qquad ----(15)$ 

The instantaneous voltage of the system and the STATCOM are independent, but the active and the reactive currents are coupled with each other through the reactance of the coupled inductor. So it is very essential to decouple the active and reactive current from each other and design the controller for tracking the required value.

#### A. Design of Current Controller

The current controller design for the above system can be done using the attempts to decouple the d and q axes equations, so that the MIMO system reduces to two independent Single Input Single Output (SISO) system. Hence, the control inputs vod and voq are configured as

$$v_{oq} = -v_{oq}^* - wL_s i_{cq} + v_{sq} \qquad ----(16)$$
  

$$v_{od} = -v_{od}^* + wL_s i_{cd} + v_{sd} \qquad ----(17)$$

The equation (18) can be obtained by replacing (14) by (16,17). Hence each row of (18) is independent of each other and thus defines an independent SISO system. Conventional frequency-domain design methods can now be directly applied for current controller. Taking the Laplace transformation of both sides of (18) and rearranging terms are given by (19) and their decoupled SISO system is shown in Fig.11.

$$\begin{bmatrix} i_{cq} \\ i_{cd} \end{bmatrix} = \begin{bmatrix} -\frac{R_s}{L_s} & 0 \\ 0 & -\frac{R_s}{L_s} \end{bmatrix} \begin{bmatrix} i_{cq} \\ i_{cd} \end{bmatrix} + \frac{1}{L_s} \begin{bmatrix} v_{oq}^* \\ v_{od}^* \end{bmatrix} \qquad ----(18)$$

$$G_q(s) = \frac{I_q(s)}{V_{oq}^*(s)} = \frac{1}{R_s + sL_s} \qquad ----(19)$$



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$$G_d(s) = \frac{I_d(s)}{V_{od}^*(s)} = \frac{1}{R_s + sL_s} \qquad ----(20)$$

For similar dynamic behaviour of the d and q – axis currents, both the d and q - axis controllers are identical and its transfer function is given in (21)

$$G_i(s) = \frac{I_{cq}(s)}{V_{oq}^*(s)} = \frac{I_{cd}(s)}{V_{od}^*(s)} = \frac{1}{R_s + sL_s} \qquad ----(21)$$
  
The transfer function of a PL controller is

$$G_{pi}(s) = K\left(1 + \frac{1}{s\tau_i}\right) = K_p + \frac{K_i}{s} - - - -(22)$$

With Kp = K,  $Ki = K/\tau i$ . The transfer function in open loop of PI controller associated with the transfer function on the a.c. system is

$$\left[G_{pi}(s), G_{i}(s)\right] = K\left[1 + \frac{1}{s\tau_{i}}\right] \left[\frac{1/R_{s}}{1 + s^{L_{s}}/R_{s}}\right] \quad - - - (23)$$

While taking  $\tau i=Ls/Rs$  and on simplification reduces to

$$[G_{pi}(s).G_i(s)] = \frac{K}{sL_s}$$
 ----(24)

The closed loop transfer function is

$$T = \frac{1}{1 + s\frac{L_s}{\kappa}} \qquad \qquad - - - -(25)$$

Thus the system behaves like a first order with an apparent time constant as

$$\tau_i = \frac{L_s}{K} \qquad \qquad ----(26)$$

The gain of K can be adjusted such a way that if it is increased too high then the system behaves as second order, otherwise responses very slow. Hence the numerical values for Kp and Ki are decided from the circuit parameters Ls and Rs from the required value of K. So the parameters of PI controller are defined

 $K_p = K_r K_i = \frac{KR_s}{L_s} \qquad ----(27)$ 

Where,  $\tau_i = \frac{L_s}{\kappa_p}$  which is taken as 0.3mseconds and with the parameters given in Table-1, value of Kpi =16.9 and Kii=3.3×103 are calculated. These parameters are used in d and q - axis current controller. The structure of the effective closed loop system is shown in Fig.12 and is replicated in both the d and q - axis current controllers.

#### **B.** Design of Voltage Controller

The relation between dc voltage  $v_{dc}$  and dc current dc  $i_{dc}$  is

$$v_{dc} = \frac{1}{C} \int i_{dc} dt \qquad \qquad ----(28)$$

The transfer function can be written as

$$G_{v}(s) = \frac{V_{dc}}{I_{dc}} = \frac{1}{sC}$$
 ----(29)

Neglecting the power loss in the source resistance and power losses in the switches, balancing the power on both sides,  $v_{sd}i_{cd} = v_{dc}i_{dc}$  ----(30)

From the above equation, we have  

$$\frac{i_{dc}}{i_{cd}} = \frac{v_{sd}}{v_{dc}} = \frac{V_s}{V_{dc}} = \frac{230}{500} = 0.46 \qquad ----(31)$$

With  $V_{dc}$  as the reference, the voltage control loop is shown in Fig.14 and it consists of inner d - axis current control loop. The active power is supplied by the d -axis current which is nothing but the ripple current of the capacitor. To make the steady state error of the voltage loop zero Proportional control is adopted here and it produces the reference d -axis current for the control of the d -axis current. The design of voltage controller is as follows:

The open loop transfer function of DC bus voltage controller is



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 $G_{op} = \frac{K^* K_{dc}}{sC} \qquad ---(32)$ The closed loop transfer function with unity feed back gain is

$$G_{cl} = \frac{1}{1 + \frac{sC}{K^* K_{dc}}} - - - - (33)$$

Where,  $\tau_v = \frac{\alpha_c}{K^* K_{dc}}$  and taking  $\tau v = 1$  msec and with the parameters of Table. I, the value of Kdc= 1.08

Then Proportional Integral controller is considering for the voltage control. Hence, the transfer function of PI controller in (22) is associated with the transfer function on dc side is

$$\begin{bmatrix} G_{v}(s) & G_{pi}(s) \end{bmatrix}_{ol} = K \left( 1 + \frac{1}{s\tau_{v}} \right) \left( \frac{1}{sC} \right) \qquad ----(34)$$
  
After taking  $\tau_{v} = C$  and on simplification

$$\left[G_{v}(s), G_{pi}(s)\right]_{ol} = K\left(\frac{1 + s\tau_{v}}{s^{2}\tau_{v}^{2}}\right) \qquad ---(35)$$
  
The transfer function in closed loop

$$\left[G_{\nu}(s)G_{pi}(s)\right]_{cl} = \left(\frac{1 + s\tau_{\nu}}{1 + s\tau_{\nu} + \frac{s^{2}\tau_{\nu}^{2}}{\kappa}}\right) \qquad - - - -(36)$$

So the system behaves like a second order system. As  $\tau_v \gg \frac{\tau_v^2}{\kappa}$  and magnitude plot in Fig.10 shows the initial slop at break point is approximately –20db/decade and hence it reduces to first order system. The value of K can be determined form root locus with approximate settling time as given in (37,38).

$$K_{vv} = \frac{K}{C} = 200 \qquad \qquad ----(38)$$

#### **III.MATLAB BLOCK DIAGRAM**



Figure 3 MATLAB model of STATCOM using variable DC voltage and constant modulation index.



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Figure 4 MATLAB model of converter gate pulse generator





### IV.STATCOM DYNAMIC PERFORMANCE WITH THE PROPOSED STRATEGIES



Figure 6 output wave form of fluctuation in distributed generation system while using distribution STATCOM



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Figure 7 output wave form of fluctuation in distributed generation system while using 48 pulse STATCOM



Figure 8 output wave form of fluctuation in distributed generation system while using distribution variable DC voltage and constant modulation index STATCOM

### V. RESULT AND DISCUSSION

In this paper, the stability of a distributed generation system connected with a variable DC voltage and constant modulation index STATCOM is studied. As shown in figure 8 this compensate the effect of voltage fluctuation within 0.5 sec and amplitude within 0.25, while using 48 pulse STATCOM figure 7 the duration was within 0.5 sec but the amplitude was high which was within 1.5-2, and when using distribution STATCOM figure 6 the duration was high 2.5-3 sec and amplitude was 0.1. Similarly as we can see from figure 6,7 and 8 that the square root of the output waveform and instantaneous flicker sensation waveforms are very smooth in figure 8 comparing to figure 6 and 7. It



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has also been shown that the interaction between a STATCOM device and a weak bus network is sensitive to non-ideal supply conditions and load variation. Therefore, tuning of the controller to achieve a fast and stable response under varying system conditions requires careful design and investigation for each specific installation. It has also been shown that the interaction between a STATCOM device and a supply system makes the rural consumers healthy and wealthy. The MATLAB results showed the applicability of the proposed methods

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