# System Reduction of Discrete Time Uncertain Model using Stability Preservation Techniques 

Ankit Sachan ${ }^{1}$, Pankaj Kumar ${ }^{1}$, Dr.Pankaj Rai ${ }^{2}$<br>Research Scholar, Dept. of EE, IIT (BHU) Varanasi, Varanasi, U.P., India ${ }^{1}$<br>Associate Professor, Dept. of EE, B.I.T. Sindri, Sindri, Jharkhand, India ${ }^{2}$


#### Abstract

An extended approach for order reduction of complex discrete uncertain systems is proposed. Using Interval arithmetic Routh Stability arrays are formed to obtained numerator and denominator of reduced order model. The developed approach preserves the stability aspect of reduced system if higher order uncertain system is stable. A numerical example is included to illustrate the proposed algorithm along with the comparison with existing techniques. An extended technique for suppressing the complexity of discrete time uncertain model using stability preservation approach is proposed. The numerator and denominator of uncertain model are suppressed by $\beta \& \alpha$ table respectively. The proposed technique guarantees the preservation of stability in reduced uncertain model and well suited in quality with other existing methods. A numerical illustration is discussed to exemplify the extended technique


KEYWORDS: Discrete Interval System, Integral Square Error (ISE), Model Order Reduction, Routh Stability Array.

## I.INTRODUCTION

In present time as the complexity of physical and geo metrical model increases many cases so it is desirable to represent reduced order system in place of its original system. System reduction for both continuous and discrete systems has been extensively studied. Conventional methods for system reduction are Aggregation method [1], Moment matching technique [2], Padé approximation [3], and factor division method [4]. Classical technique has many serious disadvantages which often lead to an unstable reduced order model even though the original system is stable. These drawbacks of classical techniques are overcome by stability preservation techniques which allows stable reduced model for stable system and having useful features such as the computational simplicity and the fitting of time-moments. However, a plenty of stability preservation techniques were developed in past. Some famous methods are Routh approximation [6], Stability equation method [7], $\alpha \beta$ table method [8], and many more.


Fig 1. Order reduction of an efficient means to enable a system-level simulation

[^0]
## International Journal of Advanced Research in Electrical, Electronics and Instrumentation Engineering

(An ISO 3297: 2007 Certified Organization)

## Vol. 4, Issue 12, December 2015

Shaked [17] for discrete multivariable system, Padè type approximations method [18] by Bandyopadhyay\&Kande, bilinear Sch warz approximation [19] by Hsieh \& Hwang etc. Apart from these, several techniques are proposed to check the novelty of reduced model.

In this brief, an extended technique is discussed for reduction of large system, where denominator is retained by $\alpha$ table of uncertain model while the numerator is obtained by $\beta$ table of the given discrete interval system, This method shows the stability of the reduced system for a stable model. The outline of the paper is organized as follows: primaries and main result of the paper are deliberated in Section II\& III, Numerical problem is illustrated in Section III with the comparison of proposed result with other existing methods. Finally, conclusions and comments are given in Section IV.

## II. PRIMALIES

A general higher order single input single output interval system of $n^{\text {th }}$ order is defined as

$$
\begin{equation*}
G_{n}(s)=\frac{\left[q_{0}^{-}, q_{0}^{+}\right]+\left[q_{1}^{-}, q_{1}^{+}\right] z+\ldots \ldots .+\left[q_{n-1}^{-}, q_{n-1}^{+}\right] z^{n-1}}{\left[p_{0}^{-}, p_{0}^{+}\right]+\left[p_{1}^{-}, p_{1}^{+}\right] z+\ldots \ldots \ldots . .+\left[p_{n}^{-}, p_{n}^{+}\right] z^{n}}=\frac{\sum\left[q_{i}^{-}, q_{i}^{+}\right] s^{i}}{\sum\left[p_{j}^{-}, p_{j}^{+}\right] s^{j}}=\frac{N_{n}(s)}{D_{n}(s)} \tag{1}
\end{equation*}
$$

where $(i=0,1, \ldots, n-1)$ and $(j=0,1,2, . ., n)$ are order of interval parameters.
and general reduced order system of $r^{\text {th }}$ order is defined as

$$
\begin{equation*}
R_{r}(s)=\frac{\left[b_{0}^{-}, b_{0}^{+}\right]+\left[b_{1}^{-}, b_{1}^{+}\right] z+\ldots \ldots \ldots+\left[b_{r-1}^{-}, b_{r-1}^{+}\right] z^{r-1}}{\left[a_{0}^{-}, a_{0}^{+}\right]+\left[a_{1}^{-}, a_{1}^{+}\right] z+\ldots \ldots \ldots \ldots \ldots+\left[a_{r}^{-}, a_{r}^{+}\right] z^{r}}=\frac{\sum\left[q_{i}^{-}, q_{i}^{+}\right] s^{i}}{\sum\left[p_{j}^{-}, p_{j}^{+}\right] s^{j}}=\frac{N_{r}(s)}{D_{r}(s)} \tag{2}
\end{equation*}
$$

where $(i=0,1 \ldots, r-1)$ and $(j=0,1, \ldots, r)$ are order of interval parameters.

## Interval Arithmetic [11, 12]

Rules based on interval arithmetic $[11,12]$ are given below. Let $[a, b]$ and $[c, d]$ be two intervals.
Addition:

$$
[a, b]+[c, d]=[a+c, b+d]
$$

Subtraction

$$
\begin{equation*}
[a, b]-[c, d]=[a-d, b-c] \tag{4}
\end{equation*}
$$

Multiplication $\quad[a, b] \times[c, d]=[\operatorname{Min}(a c, a d, b c, b d)],[\operatorname{Max}(a c, a d, b c, b d)]$
Division:

$$
\begin{equation*}
\frac{[a, b]}{[c, d]}=[a, b] \times\left[\frac{1}{d}, \frac{1}{c}\right] \tag{5}
\end{equation*}
$$

## III. MAIN RES ULT

In this section Routh approximation based $\alpha-\beta$ table method [9] is extended for discrete uncertain system. The steps to obtain reduced order system are as below:
Step 1: Bilinear transformation of higher order system by applying $z=\frac{1+w}{1-w}$
Step 2: Recip rocal transformation of higher order system is obtained as
$\hat{G}(w)=1 / w G(1 / w)$
where $\hat{G}(w)$ is reciprocal of higher order system and $G(w)$ is the original higher order system
Step 3: Formation of $\alpha$-table
The first two rows of tabulation are formed from the coefficient of denominator of $\hat{G}(w)$ and rest of entries of table is by "cross-multiplication rule" using interval arith metic rules [11, 12] such as
$\left[\alpha_{1}^{-}, \alpha_{1}^{+}\right]=\frac{\left[p_{0}^{-}, p_{0}^{+}\right]}{\left[p_{1}^{-}, p_{1}^{+}\right]}\left\{\begin{array}{l}{\left[p_{0}^{-}, p_{0}^{+}\right]\left[p_{2}^{-}, p_{2}^{+}\right] \ldots . .} \\ {\left[p_{1}^{-}, p_{1}^{+}\right]\left[p_{3}^{-}, p_{3}^{+}\right] \ldots .}\end{array}\right.$
$\left[\alpha_{2}^{-}, \alpha_{2}^{+}\right]=\frac{\left[p_{1}^{-}, p_{1}^{+}\right]}{\left[k_{1}^{-}, k_{1}^{+}\right]}\left\{\begin{array}{l}{\left[p_{1}^{-}, p_{1}^{+}\right]\left[p_{3}^{-}, p_{3}^{+}\right] \ldots . .} \\ {\left[k_{1}^{-}, k_{1}^{+}\right]\left[k_{3}^{-}, k_{3}^{+}\right] \ldots .}\end{array}\right.$

## International Journal of Advanced Research in Electrical, Electronics and Instrumentation Engineering

(An ISO 3297: 2007 Certified Organization)

## Vol. 4, Issue 12, December 2015

$\left[\alpha_{3}^{-}, \alpha_{3}^{+}\right]=\frac{\left[k_{1}^{-}, k_{1}\right]}{\left[l_{1}^{-}, l_{1}^{+}\right]}\left\{\begin{array}{c}{\left[k_{1}^{-}, k_{1}^{+}\right]\left[k_{3}^{-}, k_{3}^{+}\right] \cdots \cdots} \\ {\left[l_{1}^{-}, l_{1}^{+}\right]\left[l_{3}^{-}, l_{3}^{+}\right] \cdots .}\end{array}\right.$
where
$\left[k_{i}^{-}, k_{i}^{+}\right]=\left[p_{i+1}^{-}, p_{i+1}^{+}\right]-\left[\alpha_{1}^{-}, \alpha_{1}^{+}\right]\left[p_{i+2}^{-}, p_{i+2}^{+}\right] ; i=(1,3, \ldots, r-1)$
$\left[l_{i}^{-}, l_{i}^{+}\right]=\left[p_{i+2}^{-}, p_{i+2}^{+}\right]-\left[\alpha_{2}^{-}, \alpha_{2}^{+}\right]\left[k_{i+2}^{-}, k_{i+2}^{+}\right] ; i=(1,3, \ldots, r-3)$
$\left[m_{i}^{-}, m_{i}^{+}\right]=\left[k_{i+2}^{-}, k_{i+2}^{+}\right]-\left[\alpha_{3}^{-}, \alpha_{3}^{+}\right]\left[l_{i+2}^{-}, l_{i+2}^{+}\right] ; \quad i=(1,3, \ldots, r-5)$

Step 4: Formation of $\beta$-table
The first two rows of table are formed from the coefficient of numerator of $\hat{G}(s)$ and remaining entries are calculated from $\alpha$ table and earlier rows of $\beta$ table
Therefore
$\left[\beta_{1}^{-}, \beta_{1}^{+}\right]=\frac{\left[q_{0}^{-}, q_{0}^{+}\right]}{\left[p_{1}^{-}, p_{1}^{+}\right]}\left\{\begin{array}{l}{\left[q_{0}^{-}, q_{0}^{+}\right]\left[q_{2}^{-}, q_{2}^{+}\right] \ldots . .} \\ {\left[p_{1}^{-}, p_{1}^{+}\right]\left[p_{3}^{-}, p_{3}^{+}\right] \ldots .}\end{array}\right.$
$\left[\beta_{2}^{-}, \beta_{2}^{+}\right]=\frac{\left[q_{1}^{-}, q_{1}^{+}\right]}{\left[k_{1}^{-}, k_{1}^{+}\right]}\left\{\begin{array}{l}{\left[q_{0}^{-}, q_{0}^{+}\right]\left[q_{2}^{-}, q_{2}^{+}\right] \ldots . .} \\ {\left[k_{1}^{-}, k_{1}^{+}\right]\left[k_{3}^{-}, k_{3}^{+}\right] \ldots . .}\end{array}\right.$
$\left[\beta_{3}^{-}, \beta_{3}^{+}\right]=\frac{\left[e_{1}^{-}, e_{1}^{+}\right]}{\left[l_{1}^{-}, l_{1}^{+}\right]}\left\{\begin{array}{c}{\left[e_{1}^{-}, e_{1}^{+}\right]\left[e_{3}^{-}, e_{3}^{+}\right] \cdots \cdots} \\ {\left[l_{1}^{-}, l_{1}^{+}\right]\left[l_{3}^{-}, l_{3}^{+}\right] \cdots \cdots}\end{array}\right.$
where
$\left[e_{i}^{-}, e_{i}^{+}\right]=\left[q_{i+1}^{-}, q_{i+1}^{+}\right]-\left[\beta_{1}^{-}, \beta_{1}^{+}\right]\left[p_{i+2}^{-}, p_{i+2}^{+}\right] ; \quad i=(1,3, \ldots, r-3)$
$\left[f_{i}^{-}, f_{i}^{+}\right]=\left[p_{i+2}^{-}, p_{i+2}^{+}\right]-\left[\beta_{2}^{-}, \beta_{2}^{+}\right]\left[k_{i+2}^{-}, k_{i+2}^{+}\right] ; i=(1,3, \ldots, r-5)$
$\left[g_{i}^{-}, g_{i}^{+}\right]=\left[r_{i+2}^{-}, r_{i+2}^{+}\right]-\left[\beta_{3}^{-}, \beta_{3}^{+}\right]\left[l_{i+2}^{-}, l_{i+2}^{+}\right] ; i=(1,3, \ldots, r-7)$
$\vdots \quad \vdots \quad \vdots \quad \vdots$
Step 5: Let $A_{r}(s)$ and $B_{r}(s)$ represent the denominator and numerator respectively, of the $r^{\text {th }}$ order reduced system in general [19], i.e.
$A_{r}(s)=\left[\alpha_{r}^{-}, \alpha_{r}^{+}\right] A_{r-1}(s)+A_{r-2}(s)$
$B_{r}(s)=\left[\alpha_{r}^{-}, \alpha_{r}^{+}\right] s B_{r-1}(s)+B_{r-2}(s)+\left[\beta_{r}^{-}, \beta_{r}^{+}\right]$
Where $r=(1,2, \ldots, n)$ are obtained with $A_{-1}(s)=0, B_{-1}(s)=0, A_{0}(s)=1, B_{0}(s)=0$;
Step 6: The $r^{\text {th }}$ order reciprocal transformation system is evaluated as
$\hat{R}_{r}(s)=\frac{B_{r}(s)}{A_{r}(s)}$
Step 7: Finally reduced order system is obtained as

$$
\begin{equation*}
R_{r}(s)=1 / s \hat{R}_{r}(1 / s) \tag{11}
\end{equation*}
$$

Step 8: Inverse Bilinear transformation of reduced order system by applying $z=\frac{1-w}{1+w}$

## International Journal of Advanced Research in Electrical, Electronics and Instrumentation Engineering

(An ISO 3297: 2007 Certified Organization)

## Vol. 4, Issue 12, December 2015

## IV. NUMERICAL EXAMPLE

Consider the fourth order discrete interval system described by transfer function [27, 28]
$G(z)=\frac{[1,2] z^{2}+[3,4] z+[8,10]}{[6,6.6] z^{3}+[9,9.5] z^{2}+[4.9,5] z+[0.8,0.85]}$
Step 1: Applying bilinear transformation in higher order transfer function.
$G(w)=\frac{[5,9] w^{2}+[-18,-12] w+[12,16]}{[0.55,1.2] w^{3}+[5.9,6.65] w^{2}+[19.45,20.2] w+[20.7,21.35]}$
Step 2: After recip rocal transformation of $G(s)$
$\hat{G}(w)=\frac{[12,16] w^{2}+[-18,-12] w+[5,9]}{[20.7,21.35] w^{3}+[19.45,20.2] w^{2}+[5.9,6.65] w+[0.55,1.2]}$
Step 3: The $\alpha$ table is obtained as:

| $\alpha$-table | $[20.7,21.35]$ | $[5.9,6.65]$ |
| :--- | :--- | :--- |
|  | $[19.45,20.2]$ | $[0.55,1.2]$ |
| $[1.024,1.097]$ | $[4.583,6.0864]$ |  |
| $[3.1956,4.407]$ |  |  |

Step 4: The $\beta$ table is calculated as:

| $\beta$-table | $[12,14]$ <br> $[-18,-12]$ | $[5,9]$ |
| :--- | :--- | :--- |
| $[0.59,0.82]$ | $[6,7.5]$ |  |
| $[-2.95,-2.62]$ |  |  |

Step 5: Now transformation reduced system, using Eq. 10 is evaluated as
$\hat{R}_{2}(w)=\frac{[1.88,3.60] w+[-2.95,-2.62]}{[3.25,4.79] w^{2}+[3.19,4.40] w+1}$
Step 6: Then, reciprocal transformation of above reduced order systemusing Eq. 11 is obtained as
$R_{2}(w)=\frac{[-2.95,-2.62] w+[1.88,3.60]}{w^{2}+[3.19,4.40] w+[3.25,4.29]}$
Step 7: Applying inverse bilinear transformation in reduced order transfer function as
$R_{2}(z)=\frac{[-1.07,0.98] z+[4.5,6.55]}{[7.44,10.19] z^{2}+[4.5,7.58] z+[-0.15,2.6]}$

## V. RES ULT AND DISCUSSION

To check the superiority of the proposed method over other existing methods [27, 28] integral-square-error for reduced order models are tabulated in Table I. The integral square error is determined between transient step response of original system and its reduced order model which can be represented as
$I S E=\int_{0}^{\infty}\left[y(t)-y_{r}(t)\right]^{2} d t$
Where $y(t)$ is the step response of the original system and $y_{r}(t)$ is the step response of the original system

## International Journal of Advanced Research in Electrical, Electronics and Instrumentation Engineering

(An ISO 3297: 2007 Certified Organization)
Vol. 4, Issue 12, December 2015
TABLE I. COMPARISON OF ISE FOR REDUCED ORDER MODELS

| Model Order Reduction | Reduced Models | ISE <br> (lower) | ISE <br> (upper) |
| :--- | :---: | :---: | :---: |
| Proposed Method | $G_{2}(z)=\frac{[-1.07,0.98] z+[4.5,6.55]}{[7.44,10.19] z^{2}+[4.5,7.58] z+[-0.15,2.6]}$ | 0.0852 | 0.0377 |
| Padé\& Dominant Pole | $G_{2}(z)=\frac{[0.5921,0.6055] z+[0.8845,0.9]}{z^{2}+[0.8041,1.2465] z+[0.1437,0.3805]}$ | 0.1810 | 0.0 .741 |
| Retention Method $[27]$ | $G_{2}(z)=\frac{[-1.328,1] z+[3.522,5.85]}{[6.89,8.14] z^{2}+[3.94,5.44] z+[0.55,1.8]}$ | 1.1292 | 0.0443 |
| $\gamma-\delta \operatorname{method}[28]$ |  |  |  |



Fig 1(a). Step response of Original and reduced systems using Kharitonov's theorem


Fig 1(b). Step response of Original and reduced systems using Kharitonov's.theorem
In the fig 1(a) \& fig 1(b), it shows the step response of original system and reduced order systems using Kharitonov's theorem in which graph is plotted for amplitude Vs time (seconds). Throughput is the average rate of successful message delivery over a communication channel. It is observed from fig 1(a) \& fig 1(b) that step responses of original system and reduced order interval system using proposed method are close to each other which show the accuracy of proposed method.

## International Journal of Advanced Research in Electrical, Electronics and Instrumentation Engineering

(An ISO 3297: 2007 Certified Organization)

## Vol. 4, Issue 12, December 2015



Fig 2(a). Frequency response of Original and reduced systems using Kharitonov's theorem


Fig 2(b). Frequency response of Original and reduced systems using Kharitonov's theorem
In the fig 2(a) \& fig 2(b), it shows the frequency response of higher order system and lower order systems using Kharitonov's theorem in which graph is shown for magnitude (dB) \& phase (deg) Vs time (seconds).

## VI.CONCLUSION

This paper presents an extension of the $\alpha-\beta$ table technique [13] for reduction of higher order discrete time interval systems. The numerator polynomial and denominator polynomial is obtained by using $\beta$ table and $\alpha$ table respectively. The developed technique is conceptually easy and preserves the accuracy and stability of reduced order model if the higher order system is stable. The proposed method produces lesser values of error indices when compared with other existing methods [27,28]. A numerical example is discussed and compared with some existing techniques.

## References

[1] M. Aoki, "Control of large-scale dynamic systems by aggregat ion,"IEEE Trans. Aut omat. Contr., vol. AC-13, pp. 246-253, 1968.
[2] N. K. Sinha and B.Kuszta, " Modeling and identification of dynamic systems," New York: Van Nostand Reinhold, pp.133-163, 1983
[3] Y. Shamash, "Stable reduced order models using Padètype approximation," IEEE Trans. Automat. Contr., vol. AC-19, pp. 615-616, 1974.
[4] T.N. Lucas, Fact or division: a useful algorithm in model reduction, IEE Proc. 130, Vol. 6, 362-364, November 1983.
[5] C. F. Cfen and L.S. Shich, "Anovel approach to Linear model simplificatiion", Int. 3. Contr., Vol 8. pp.561-570,1968.

## International Journal of Advanced Research in Electrical, Electronics and Instrumentation Engineering

## (An ISO 3297: 2007 Certified Organization)

## Vol. 4, Issue 12, December 2015

[6] V. Krishnamurthy and V. Seshadri, "Model reduction using the Routh Stability criterion,"IEEE Trans. Automatic Control, vol. AC-23, pp.729731, Aug. 1978.
[7] T.C. Chen, C.Y. Chang, and K.W. Han, "Stable reduced-order Padè approximants usingstability equation method," Electron. Lett., 1980,16, pp. 345-346.
[8] M.F. Hutton and B. Friedland, "Routh Approximation for Reducing Order of Linear Time Invariant System", IEEE Trans. Autom. Control,20, 329-337, 1975
[9] R.K. Appiah,"Padè methods of Hurwitz polynomial approximation with application to linear system reduction," Int. J. Control, 1979, 29,pp. 39-48.
[10] K. Glover, "All Optimal Hankel-norm Approximations of Linear Multivariable Systems andtheir L $\infty$ Error Bounds," International Journal of Control, pp. 1115-1 193, vol. 39 (6), 1984.
[11] E. Hansen, "Interval arithmetic in matrix computations Part I," SIAMJ. Numerical Anal., pp. 308-320, 1965
[12] E. Hansen and R. Smith, "Interval arithmetic in matrix computation.Part II," SIAM J.Numerical Anal., pp. 1-9, 1967
[13] M. Sharma, A. Sachan, D. Kumar, "Order reduction of higher order interval systems by stability preservation approach," inPower, Control and Embedded Systems (ICPCES), 2014 International Conference on, vol., no., pp.1-6, 26-28 Dec. 2014
[14] V. L. Kharitonov ,"Asymptotic stability of an equilibrium position of a family of systems of linear differential equations," Differentsial'nye Uravneniya, 14, pp.2086-2088, 1978.
[15] R. Barmish, "A generalization of Kharitonov's four polynomials concept for robust stability problems with linearly dependent coefficient perturbation," IEEE Tran. Auto. Cont., vol. 31, No. 2, pp. 157-165, 1989.
[16] Y. Shamash, and D. Feinmesser, 'Reduction of discrete time systems using a modified Routh array', International Journal of Sy stem Science, vol. 9, no. 1, pp. 53-64, 1978.
[17] Y. Bistritz;U. Shaked.".Discrete multivariable system approximations by minimal Pade-typestable models". IEEE Transactions on Circuits and Systems,Volume: 31 ,Issue: 4 ,Page(s):382-390,1984.
[18] P. Parthasarthy and K. N. Jayasimha, 'Modeling of Linear Discrete-Time Systems usingModified Cauer Continued Fraction'. Journal of Franklin Inst, vol. 316, no. 1, pp. 79, 1983.
[19] Bandyopadhyay and Kande (1988) proposed a method of model reduction for discrete-timecontrol systems by Padetype approximations.
[20] Y. Choo, 'A Note on Discrete Interval System Reduction via Retention of Dominant Poles',International Journal of Control, Automation, and Systems, vol. 5, no. 2, pp. 208-211, 2007.
[21] N. Pappa and T. Babu, 'Biased model reduction of discrete interval system by differentiationtechnique', Annual IEEE India Conference, (INDICON), pp. 258-261, 2008.
[22] B. Vishwakarma, and R. Prasad, 'Clustering method for reducing order of linear systemusing Pade approximation', IETE Journal of Research, vol. 54, no. 5, pp. 326-330, 2008.
[23] S. K. Mittal, and D. Chandra, 'Stable optimal model reduction of linear discrete time systemsvia integral squared error minimization: computeraided approach', Jr. of Advanced Modelingand Optimizat ion, vol. 11, no. 4, pp. 531-547, 2009.
[24] V. P. Singh and D. Chandra, 'Model reduction of discrete interval System using dominantpoles retention and direct series expansion method', The 5th International Power Engineering and Optimization Conference (PEOCO), pp. 27-30, 2011.
[25] Hwang, C., and Hsieh, C.S., "Order reduction discrete time system viz., Bilinear RouthApproximation", ASME J. Dyn, Syst, Meas., control 112, PP 292 - 297, 1990.
[26] M. F. Hutton, "Routh approximation method for high-order linear systems," Ph. D. dissertation, Polytechnic Inst. New York, Brooklyn, N.Y., June 1974.
[27] O. Ismail, B. Bandyopadhyay, and R. Gorez, "Discrete interval system reduction using Padé approximation to allow retention of dominant poles," IEEE Trans on Circuits and Systems-I: Fundamental theory and applications, vol. 44, no. 11, pp. 1075-1078, 1997.
[28] A. K. Choudhary and S. K. Nagar, "Gamma Delta Approximation for Reduction of Discrete Interval System," International Joint Conferences on ARTCom and ARTE


[^0]:    Developments made in the direction of downsizing the complexity of an uncertain system after the idea of solving the arithmetic of uncertain system by Hansen \& Smith [11,12]. However, Practical systems like cold rolling mill, stirred tank reactor, oblique aircraft and electric motors, contains disturbance in model dynamics due to sensor noises, actuator constrains, etc. which are most suitably represented by continuous or discrete type uncertain models, instead of deterministic mathematical models. Many methodologies based on discrete time uncertain system came into picture such as Routh array method by Shamash \& Feinmesser [16], minimal Padè type stability method by Bistritz and

